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A Miniature Deep Sea Temperature Data Recorder: Design, Construction, and Use

by

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Technical Report

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ABSTRACT

A miniature temperature recorder has been developed to be used with the hydraulic piston sediment corer (HPC) on the Deep Sea Drilling Project (DSDP). The instrumentation fits into pressure-sealed slots in the wall of the HPC, allowing temperature measurements to be made simultaneously with coring operations. Temperatures from -2 to 70°C are measured to a resolution of about 0.01°C. Up to 1300 13-bit measurements are recorded in random access memory (RAM), at a sampling rate ranging between 0.1 s to over 100 min., as specified by the operator in a program loaded into a microprocessor of the instrument. During recording the instrumentation uses about 3.5 mamp at 7.5 volts, which can be supplied for about 20 hours of operation by a custom-made pack of silver-oxide batteries. The corer is normally left motionless in the sediment for about 10 min. to allow extrapolation of the measured temperatures to equilibrium in-situ temperature. Examples of data from DSDP Leg 86 are given.

I. INTRODUCTION

Marine geothermal measurements require that the vertical temperature gradients in sediments be determined accurately. This is usually accomplished by instrumentation lowered to the sea floor from an oceanographic ship or drilling platform. Probes and cores used with oceanographic vessels are typically designed to penetrate the bottom from 1 to 10 m, with the gradient measured by 4-7 thermal probes rigidly attached along the probe. In deep-sea drilling, instrumentation has been used to measure temperatures at intervals during drilling frequently to depths of 200-300 m (Erickson et al., 1975; Yokota et al., 1980), and occasionally to 600 m below the sea floor (Erickson and Von Herzen, 1978).

The much greater penetration achieved by drilling techniques provide the opportunity to investigate variability of heat flow with depth. Although many results indicate relatively constant heat flow with depth (Erickson et al., 1975; Hyndman et al., 1985), as expected for steady-state conductive thermal transfer, oceanographic studies suggest that some regions may have more complicated thermal signatures. On or near continental margins, processes such as rapid sedimentation or slumping, or bottom water temperature variations, may produce more complex or disturbed thermal profiles. Hydrothermal circulation in ocean crust is a process which dominates heat transfer over much of the youngest sea floor, and extends to sea floor as old as 80 m.y. in some ocean basins (Anderson et al., 1977). Recently it has been found that this mechanism may involve slow fluid transfer through the overlying sediments (Anderson et al., 1979; Von Herzen et al., 1982; Becker and Von Herzen, 1982). The vertical component of interstitial fluid flow will produce a distinctive temperature profile with depth different from conductive profiles, essentially a exponential curve with a depth scale which depends primarily on the rate of vertical flow. Even extremely slow rates of pore fluid flow (~ 10^{-8} cm sec⁻¹ = 0.3 cm/yr) have important ramifications for schemes to dispose of wastes in the sea floor, esp. radioactive wastes which must be isolated for 10^{5} yrs or more.

Successful in-situ measurements of temperature in Deep Sea Drilling Project (DSDP) holes began with Leg 5 in 1969. Over the past decade, measurements have been obtained at only a few depths downhole in a small fraction of the total number of holes drilled by DSDP, due to several factors:

1) Drilling objectives did not justify the additional time required by such measurements.

2) The harsh environment of the drilling operations, primarily large shocks and accelerations, rendered downhole instrumentation inoperative (this factor has been substantially reduced by development of all solid-state instrumentation (e.g. Yokota et al., 1980)).

3) Sediment physical properties did not allow measurement of equilibrium temperatures (either too soft to hold the measurement probe steady, or too hard to penetrate).

4) Sediments are thermally disturbed due to drilling.

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The successful development of a DSDP hydraulic piston corer (HPC) has allowed retrieval of remarkably undisturbed sediment cores to depths of 200 to 300 m beneath the sea floor. The HPC is a nearly ideal vehicle for an in-situ temperature sensor because it deploys well below (up to 10 m) the drilled and thermally disturbed hole for core retrieval (Fig. 1). Furthermore, its almost continuous deployment for the upper several hundred meters of sediments on many DSDP holes provides the opportunity to obtain detailed depth profiles of temperature with little or no cost in additional ship time.

Here we describe, in detail, instrumentation developed to obtain in-situ temperature during deployments of the HPC, along with examples of data. The instrumentation was first used on DSDP Leg 86 in 1982 (Horai and Von Herzen, 1985) and extended through the drilling program carried out on D/V GLOMAR CHALLENGER. Its use presently (January 1986) continues with the Ocean Drilling Program from D/V JOIDES RESOLUTION.

II. DESIGN CONSIDERATIONS

The HPC (hydraulic piston corer) is up to 10 m long and about 9 cm diameter (Anonymous, 1984). The most useful location of the temperature sensor is in the cutting shoe (the tip of the coring tool) where it is farthest from the temperature disturbances of the drill bit. Also, the tip of the coring tool is thinner than the rest of the core barrel, which suggests that it takes less time to reach equilibrium temperature with the sediment. The temperature recorder electronics could be anywhere, as long as it did not impede the operation of the coring tool.

Initially, we considered putting the electronics in a hollow (doughnut shaped) cylinder at the top of the coring tool, and running the sensor wires down the length of the coring tool. Although the users were willing to give up 15 cm of coring length for the electronics at the top of the core barrel, we could not determine how to run the sensor wires past the threaded sections of the coring tool to the sensor in the tip.

The cutting shoe has a 1.43 cm thick wall. If a portion of that wall could be hollowed out to make room for the electronics, then the sensor wires could be connected directly to the temperature recorder. If the electronics were contained on a hybrid circuit, they could be put into a space no larger than an ordinary large integrated circuit package. The \$8000 for the hybrid circuit layout and masks plus \$1800 for each unit produced were reasonable costs.

Engineers at DSDP designed as large a cavity as possible in the coring shoe to hold a very small electronics package and withstand the 10,000 psi pressure at the bottom of the 21,000 ft drill string. The modified coring shoe evolved in steps. Initially, the cavity was to be covered by a lid on the outside of the coring tool, but the possibility existed for loosening of the cap and jamming inside the drill string. The second version had the top and the bottom of the coring shoe separate, exposing the cavity for the electronics. The two sections were bolted together and O-ring seals kept the water and pressure out of the cavity. The strength of the bolts was



Figure 1. Schematic configuration of drill ship deploying HPC during coring operations

determined as insufficient to withstand the side forces the coring shoe could encounter. The third version replaced the bolted fastening with a threaded section. The strongest heat treatable steel available was used for the shoe, so that the cavity could be as wide as possible.(See Appendix A for the calculation of the strength of the cavity walls.)

The instrumentation was designed to record temperatures up to 70°C. It also had to survive the shocks encountered during the traverses of the drill string at high speed (200 m/min), which we estimated at approximately 5000 g (equivalent to a free-fall of 5 ft. stopped in 0.01 in.). The hybrid could withstand these shocks, and the rest of the printed circuit board was potted to keep the parts from moving and breaking off the printed circuit board.

III. THE INSTRUMENT

A. DESCRIPTION

The temperature recorder can make and store 1300 temperature measurements and occupies 16 cm^3 (1 in.³). The battery was designed to last from 35 to 50 hours (active, the lesser amount when cold) to 400 hours (data stored mode). It occupies only 10 cm³ (0.6 in.³) (see Fig. 2). Initially, the recorder is coupled to a small computer and its program is loaded each time before use. Up to 8 time delays and recording rates can be inserted, although only one is normally used.

After the program is loaded, and the recording parameters selected, the two halves of the coring shoe are screwed together (Fig. 3). The electronics and battery package are sealed from ocean pressures and water by O-ring seals. The coring shoe is taken to the drilling platform where it is attached to the bottom of the hydraulically driven piston corer. A tilt switch within the battery package activates the timing of the recorder when the shoe and coring tool are lifted to vertical (with the cutting edge pointing down).

The corer is lowered to the bottom of the drill string, driven into the sediment, and held there for about 5 to 10 minutes. The corer is pulled to the top of the drill string and the core and core shoe are removed. In the lab the coring shoe is opened, and a plug from the computer connected to the temperature recorder. The data are read into the computer over an RS-232C (teletype) line. The computer stores the 1300 temperature points on tape or a disk. The data, which are in raw form, can be converted to actual temperatures and stored or plotted. Also, an algorithm to determine the in situ bottom core temperature measurements. A personal computer used to communicate with the temperature recorder is less expensive than constructing a special unit to store and process the data.

B. DESIGN

The temperature recorder block diagram is shown in Figure 4. A thermistor in a Wheatstone bridge provides voltage for an analog-to-digital

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Figure 2. Photograph of temperature recorder and battery. The body of the recorder is 12 cm long.



Figure 3. Photograph of instrumentation and battery pack outside of special hydraulic piston coring (HPC) cutting shoe (2 pieces) designed to contain them.



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Figure 4. Block diagram of HPC temperature recorder.

(A/D) converter. An RCA 1802 microprocessor is used to couple the output of the A/D converter to random access memory (RAM). The microprocessor also controls when the A/D converter takes a measurement. There are 3k bytes of memory; about 0.5k are used for the program and the rest is used to store data. A universal asynchronous receiver transmitter (UART) is used to couple the computer to a teletype type output and input.

The input instructions are transferred through the UART to the computer. The program and answer to inquiries are transferred from the computer through the UART to the microprocessor and its memory.

A voltage regulator provides the +5 volts for the digital and analog circuitry. A voltage inverter provides -5 volts used only by the A/D converter. Five silver oxide hearing aid batteries provide 8 volts to the voltage regulator. To save energy, the voltage to the A/D converter is switched on only when it is being used. To save additional energy, when the data memory is filled, the voltage is dropped to 4 volts, and the microprocessor halts.

After the recorder is brought back to the surface, the coring shoe is opened up in the lab to gain access to the recorder electronics. The recorder is plugged into the on-deck computer, the recorder voltage is raised to 5 volts and the microprocessor on the recorder is started up again. The internal program then dumps the data out, stops the microprocessor and lowers the recorder voltage to 4 volts when completed. Because of the low standby power drain, the battery does not have to be disconnected. In operation at sea, the entire unit has to be removed each time so that the O-ring seals and the threaded opening to the electronics can be cleaned of mud, grease, and any dirt which may have accumulated during the deployment.

C. CONSTRUCTION

The circuitry had to be made quite small in order to fit into the small space available. A hybrid circuit is very small in size, and can be made relatively economically in small quantities. The first question for a hybrid circuit is determining whether it will be a thick or thin film hybrid. All hybrids are made on a ceramic substrate, typically .025 in. thick. A thin film hybrid is a single layer of gold lines placed on the ceramic through a mask. Since such a technique does not allow for wiring crossovers, the circuit cannot be very complex. A thick film hybrid consists of alternate layers of gold lines and glass insulator on a ceramic substrate. Each layer is silk screened on and the pattern is fired before the next layer is applied. Interconnections between the wires on different layers are made through holes ("vias") in the insulating layers. Because of the multiple layers, multiple cross-overs are possible to accommodate a more complex circuit. The complexity of this circuit, which used 20 integrated circuits, required a thick film hybrid circuit.

The second problem was finding a hybrid manufacturer who was willing to manufacture only a few (5-10) units. There was a fairly large engineering effort in laying out the circuit, and sources for just a few of each type of integrated circuit had to be found. Larger hybrid companies preferred to use their limited engineering expertise on products with a larger production. Even the smaller company we found (Transistor Specialities, Inc, Danvers, Mass.) accepted the job with reluctance, but then was determined to do the job, even though it was more complicated (5 metalization layers) than his colleagues in other companies thought prudent.

The design at first incorported two 40-pin packages (0.6 in. spacing between pin rows, 2 in. long) to hold the circuitry. Later the package manufacturer came out with a 3-in.-long package with pins on 0.6 in. spacing. It was long enough to hold the whole circuit, except for certain large parts. This simplified the printed circuit, since no connections would now be needed between two hybrid packages. We were fortunate that someone else underwrote the tooling for this larger package, as a die designed for a new package costs around \$14K. Additional costs would be incurred for the lid die and the solder preform die.

The circuit had to be designed for integrated circuits that are available in dice form. Not only does the manufacturer need to have them available in dice form, but the dice distributer must be willing to stock the part. They are not very interested if the part is not in great enough use to allow them to sell the minimum order from the manufacturer. For RCA this minimum is one wafer of dice. For others it might be 100 pieces, even though the piece costs \$100 each. Japanese manufacturers did not seem to be interested in selling dice at all. Buying small numbers of dice is a problem. The designer must consider what is available in small numbers, or be willing to purchase more dice than needed. We even designed for one part that later was discontinued at the time of purchase. Fortunately, the hybrid had been inadvertently layed out for a larger equivalent part that was still available. Also, lead times were long, and slower moving items are not manufactured frequently.

After the hybrid is manufactured, it must be tested without physically disturbing the circuit very much. A single bond wire can be gently raised from the substrate with only a modest chance of ruining the IC; the entire IC can be replaced with a greater chance of ruining the whole hybrid. Generally only the hybrid manufacturer can electrically probe the hybrid. It was not difficult to determine if the circuit was working incorrectly. Ingenuity was required to devise tests to determine what part was at fault, without removing any of the parts.

Ultimately, we solved all the problems, but it was considerably more difficult than working with a printed circuit. The circuit should be designed so that faults can be isolated. The most difficult problem is designing a technique for determining which of the several ICs fastened to the same lead is at fault.

D. PRINTED CIRCUIT LAYOUT

We wanted to put all the components inside the hybrid in order to minimize the size of the package. However, a few parts were left out of the hybrid either because they were too large to fit, there was insufficient space in the hybrid package for them, or we wanted to be able to change them





after the hybrid was built. Means were needed to connect the hybrid pins to these parts, to the temperature sensor and to the connectors for power and communication. The usual method for interconnecting parts, a printed circuit, was used (Fig. 5). It was unusual in that it was very thin, only 1/64 in. thick, in order to use as little space as possible.

The bridge components and the A/D integrating and zero offset capacitors were placed at one end of the printed circuit. The hybrid package was placed in the middle. A thin (.008 in.) insulating sheet of fiberglass-epoxy board was placed between the metal hybrid package and the printed circuit. On the other end were the tantalum bypass capacitors, crystal; voltage regulator resistors, A/D frequency resistor, the data/control connector, and the battery connector.

Initially, the circuit runs were wide (0.32 in.) and the spaces between them thin (.010 in.), with the idea that the lines could be robust and the spaces between the lines did not have to be very strong. This spacing proved to be too close. The etching was not complete in several places, and fine hairs of solder from the edges of the traces could easily bridge the gaps between the lines. After an initial run of several instruments, the circuit board was layed out again to accommodate some changes, and the opportunity was used to make the traces and spaces between them equal. The board was changed to accomodate some smaller packaged bridge resistors (two to a package instead of one), an external voltage regulator for the bridge, and centering of both the large integrating capacitor and the communication connector. The centering made use of the larger clearance available in the center of the kidney-shaped hole into which the package was placed (Fig. 6). Initially, it was supposed that the thin board would bend, and the mold would force the taller components at the edge of the board inside the confines of the space available. The narrow board was quite stiff, and would not bend much. This stiffness was actually a blessing, because this board was all that really provided any stiffness to the package. This stiff board made an especially good method to secure the connector to the rest of the circuitry without flexing its soldered connections.

All parts leads were bent over parallel to the board to maximize packing efficiency. Even so, some of the leads were .025 in. thick, about 10% of the available height. The bending of leads also provided for a stronger connection, since the board was so very thin.

E. POTTING

Very large shocks to the electronics package were anticipated as it travelled down the drill pipe. Similar problems had been solved by enclosing everything in a semi-rigid potting material. The rigidity of the potting material kept the parts from being ripped from place by their own inertia during shock. Some flexibility is desirable to prevent the parts from being broken or leads sheared from the slight differences in thermal expansion between the potting material and the items being potted. A reported high failure rate of hermetically sealed tantalum capacitors of the 150D series caused us to examine more closely several of these capacitors.



Figure 6. Sectional drawing of temperature instrumentation deployed in slot in wall of HPC. Dimensions given in inches.

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They appeared fragile with a gossamer appearance, so solid tantalum capacitors potted in epoxy were used.

Several potting materials were tried. A very hard epoxy chipped and flaked off in the thin sections, although it proved satisfactory for the battery packs, where the epoxy sections were thicker. All the slightly flexible epoxies and polyurethanes became softer and much more flexible at 70°C. A polyurethane from CONAP (Olean, N.Y.), EN-9, was selected for this service. It too became quite flexible at 70°C, but it remained tough.

Creating a suitable mold was one of the most difficult tasks. In an attempt to save money, we tried to take an impression of the cavity into which the electronics were to fit. We got a rubber impression, but it was flexible. Eventually, we got a hard epoxy impression, from which a hard polyurethane mold was made. Several electronic packages were cast which had to be sanded and filed to enable them to fit into the cavity.

Later, we received a sample cavity made of steel from the DSDP in which our molded units could be tested for fit. This test fixture was converted into a two-piece mold by cutting it in half, and replacing all but .010 inches of the cut with a shim. This solved our molding problems. A hard mold made by a mold-making shop would have cost \$1,200, which in retrospect would have been less expensive than our mistakes.

The instrument casting was made by filling the mold from the bottom, with the mold in a vertical position, which reduced the possibility of entrapped air. Because of the thin covering and bubble-prone potting techniques, the whole circuit was thinly coated with a polyurethane material (Laminar X500; Midland Div., The Dexter Corp., Hayward, CA) designed for that purpose. This would keep the circuit dry, protecting it from condensation as the package cooled off, and from accidental sea water exposure while opening up the package to read out the data. Another problem was solvent evaporating from the circuit board coating. The drying time recommended by the manufacturer of the printed circuit board coating was long enough for the coating to harden, but not long enough for all the solvent to have evaporated. The first board coating that we used worked, but we changed in mid-stream to a second type, which had the extra solvent.

The thermistor used to measure temperatures is Part No. GB42JM63 (Fenwall Electronics, Framingham, MA) with a nominal resistance of 15k ohms at 25°C. It is potted within a small (0.120 in. O.D.) tube with a high thermal conductivity silicone rubber (Eccosil 4952, Emerson and Cumming, Canton, MA). This probe is then potted in proper position with the electronics package to comprise a rigid unit.

The battery potting is done with a hard epoxy, (Stycast 2651-MM, Emerson and Cumming Co., Canton, MA), as the batteries could withstand the shrinkage forces of the setting epoxy. We first used a fairly hard rubber mold, filled from the top. A short cable with a small connector on it is used to power the electronics module. The main problem is keeping the epoxy from running out along the cable, making it stiff. The cable is given an accordian fold, to accomodate slight differences in cable length. The lack of stiffness in the mold was compensated for by making the package slightly thinner than the cavity. Thus clearance could exist where the battery protruded slightly into the rubber mold. Later, the steel mold for the electronics was plugged to a length of 3 in., and used to mold the batteries.

F. BATTERIES

A cavity for the batteries in the coring shoe was placed circumferentially, rather than axially, adjacent to the electronics cavity, to minimize the length of the coring shoe. The corer takes a 10 m long (maximum) core and we did not want to use any more space than necessary. A cavity with the same cross section as for the electronics was used for the battery, even though it could have been a little narrower. The 3 in. length accomodated the 5 button cells. The Eveready S76E silver oxide button cell for hearing aids was chosen for the largest cell that would fit into the space available. The cell is 0.455 in. dia. x 0.211 in. thick. A battery assembly company (Brentronics, Commack, NY) welded 5 of these together for us, with the cells side by side.

The cells have 190 mamp-hr capacity at 25°C, and can operate at 70°C if not held at that temperature for too long. The electronics uses only 3.5 mamp maximum when the A/D converter is running, which should give 50 hours of life. This is more than sufficient for perhaps 20 each or more deployments to the bottom for coring and return (average time 2 hrs. each). The electronics package is opened up each time, and the battery pack can be easily replaced at the end of its life. Even longer life is achieved in the "computer running only" state with a current drain of 1.3 mamp for 140 hours (about 6 days), or in the standby mode (memory retained, computer idling) with a current of 0.4 mamp for a 475 hour life (20 days).

The operating time with the A/D running, based on a 190 mamp-hr battery capacity at 25°C, is reduced to about 35 hrs (70% capacity) at 0°C. However, the battery pack was exhausted after only 20 hours at 0°C, with a recorder current measured at 3.5 mamp. This was verified in the laboratory by connecting a resistor to the battery pack which would draw 3.5 mamp at 7.5 volts (the nominal battery voltage, which is nearly constant during discharge). The additional reduction to 20 hours is unexplained. It could be caused by old batteries, which have reduced life, or by the heat from welding tabs to such small cells. Fortunately, 20 hours is still sufficient for 8 to 10 measurements.

The battery pack includes an inexpensive tilt sensor which commences timing when the corer is lifted to the vertical position. Originally it had been placed in the electronics package, but there was insufficient space for it, and it had to be put into the battery pack.

A small 4-pin connector was used to connect the battery pack to the electronics. The cable part was fastened to the disposable battery, to eliminate any cable fatigue in the permanent electronics.

IV. CALIBRATION

A. TEMPERATURE CALIBRATION

The temperature recorders are calibrated in a stable temperature controlled bath from 0 to 60°C at each 10°C intervals. The bath temperature is measured with a platinum resistance thermometer and bridge while the temperature recorder output is monitored with a computer. The resistance of the thermistor in the temperature recorder at each temperature point is calculated and used to fit to the thermistor equation (see below). The parameters of the thermistor equation for a particular recorder are used for calibration of the data acquired by that particular temperature recorder.

1. TEMPERATURE BATH TECHNIQUE

A stable temperature bath is used to control the temperature of the temperature recorders. We used a Tronac (Orem, Utah) bath, with a stability of about 2 m°C or better. Each temperature recorder is put into the finger of a waterproof surgeons glove and connected to a cable for plugging into the interface box. The glove and recorders are taped to a rigid plastic bar and inserted as far into the bath as the top of the glove will safely allow. A plastic bar is used because of its low heat conductivity and heat capacity. A separate ground wire is placed in the bath and connected to the ground connection (black binding post) on the interface box. This shorts out the 50 volts ac that would otherwise be in the bath and couple noise into the recorder. We calibrate all the recorders that are on hand at one time, to save the time it takes to run through the temperature range with the bath. Each recorder cable is plugged into the interface box at each temperature. Although we have not used it, a switch box would be desirable.

We started at the lowest temperature $(0^{\circ}C)$ and worked up, because the bath can be heated more quickly than it can be cooled. The bath temperature was measured with a laboratory standard platinum resistance thermometer and a precision resistance bridge. Calibration of the platinum resistance thermometer is maintained in the usual way with a triple point of water cell at 0.010°C. The two 10 turn potentiometers (coarse and fine) used to set the bath temperature on the controller have been replaced with a stable resistance box, to enable return to a given temperature setting.

The interface box was connected to an HP 85 desktop computer which both down-loaded the RCA 1802 A/D Read program into the recorder and read the recorder output and listed it on the screen. The HP 85 program, which read the recorder output, also had an option to average 50 readings, to get a better estimate of the measurement, which usually jumped between 2 and occasionally 3 least significant bits. The output of the interface box is in hexadecimal numbers in ASCII on an RS232-C line, so that any computer can be used.

2. LINEARITY AND SENSITIVITY

The readings from the recorder are used to determine the resistance of the thermistor at each temperature calibration point by using the



Figure 7. Thermistor bridge divider ratio and sensitivity vs. temperature.

parameters of the .01% accuracy resistors used in the bridge circuit for the thermistor (Fig. D-1, Appendix D). The output of a resistance bridge with a thermistor in one leg is fairly linear in temperature (see Fig. 7). This linearity keeps the sensitivity fairly uniform over the range, varying from 8 to 11 m°C/bit for this recorder range of -2 to 70°C. These resistance values vs. temperature are used in a least squares best fit to the Steinhart and Hart equation for thermistors. The parameters of this thermistor equation are then used to convert the raw data measured by the recorder into temperature. The thermistors used in the recorders are matched to within 0.25°C over the 0 to 70°C temperature range. To optimize temperature accuracy, the thermistor calibration coefficients for each individual recorder must be used in data conversion.

The temperature recorders should be recalibrated often enough to keep any changes from affecting the measurements. Some calibration shifts of up to 50 m°C were found in 2 instruments after one year (see below). The problem was that we do not know when the calibration changed, so we do not know which data were affected. Hence, a method to check the calibration at sea is desirable. In these cases, only the calibration offset changed, not the slope, so perhaps a single point calibration would be sufficient while at sea to assure the validity of the data.

B. THERMISTOR CURVE FITTING TO THERMISTOR CALIBRATIONS

We have found that the thermistor calibrations are fit very well (to better than a few millidegrees over a 60° range) by a 3 parameter equation (Steinhart and Hart, 1968)

 $T^{-1} = A + B \ln R + C (\ln R)^3$

where T = temperature (K)

R = thermistor resistance (ohms)

A, B, C are constants, different for each thermistor. The constants A, B, C are determined from a least squares best fit to thermistor calibration data as explained in greater detail in Appendix E.

C. THERMAL RESPONSE TESTS OF SPECIAL HPC CORING SHOE (DSDP)

The following describe tests of the thermal response of the shoe, as made in the laboratory at WHOI (January 1982) using various materials and procedures. A calibrated thermistor was placed in a 1/8 in. outer-diam. steel tube, which was in turn inserted in the mating hole in the shoe designed to contain the temperature sensor. The thermistor was located about 1-1/2 in. from the cutting edge (we did not attempt to vary this distance during these tests). The thermistor was connected to a signal processor and data logger (actually a breadboard of the circuitry employed in the in-situ recording system). Thermal responses were obtained in 3 different materials:

- 1) in a well-stirred water bath
- 2) in gelatin
- 3) in clay (potter's clay).

All of these tests were carried out in a bucket of the respective materials about 10 in. diam. x 12 in. deep. The shoe was inserted into the middle of the materials starting at time t = 0. The results in Fig. 8 shows the fraction (F) remaining of the total temperature change for each experiment, which is calculated as $F = \frac{T_{\infty} - T(t)}{T_{\infty} - To}$

where To = HPC shoe temperature before starting test

- Τœ = temperature of medium (or bath) in which HPC shoe is inserted
- T(t) = HPC shoe temperature at time t after start of test.

The curves in Fig. 8 show that the HPC shoe will decay to about 20-25% of its initial temperature anomaly 5 minutes after penetration. An exact determination of in-situ temperatures requires fitting of the thermal response curve to a theoretical (numerical) model (see Sec. VI.). The thermal response curves have some peculiarities which may be explained by the experimental conditions:

1. The thermal decay curve in the well-stirred water bath has an inflection between 20-30 sec when plotted on a logarithmic time scale. This may represent the thermal response of the shoe geometry. On the other hand, it may be the effect of the test conditions. The water outside of the HPC cutter is well-stirred whereas the water inside the shoe was mixed only by raising and lowering the shoe a few times during the test. This was probably first done at about 20-30 sec. after beginning this test.

2. The thermal response of the HPC in gelatin appears slightly faster than in clay, whereas the thermal conductivity of the gelatin (1.48 mcal°C⁻¹cm⁻¹sec⁻¹, or TCU) is less than half that of the clay (3.65 TCU). We surmise that there may have been a thin layer of air between the HPC cutter and the clay, caused by the way in which the shoe was forced into the clay. The relatively high shear strength of the clay caused the level of clay inside the shoe to be much less than the level outside (3-7/8 in. vs. 7-3/4 in., respectively, from the shoe cutting edge) during the test. whereas these level differences during the tests with gelatin were observed to be much less. (This problem would not be expected during normal use of the HPC in-situ due to the high hydrostatic pressure.)

Another possibility is that T_{∞} in the gelatin (18.8°C) was measured incorrectly. It was significantly lower than that of the clay even though all tests were done in the same laboratory. It was noted after the tests that the gelatin rapidly disappeared as a result of dehydration (evaporation), perhaps causing the surface of the gelatin to have a lower temperature than the interior of the bucket used in the tests.



Figure 8. Thermal response of the instrumentation within the HPC shoe in laboratory tests with different types of materials.

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V. PROGRAM LOADING AND DATA HANDLING

The temperature recorder, contained within the wall of the coring shoe, can make up to about 1300 temperature measurements during coring operations, consisting of: 1) the descent of the coring tool to the bottom of the drill hole, 2) the insertion of the coring tool in the sediment (the temperature measurements that we want), and 3) the ascent of the coring tool to the surface. Temperature can be measured over the range from $-2^{\circ}C$ to $+70^{\circ}C$, with an average resolution of 9 m°C. A time delay can be set so that the temperature recorder is near the bottom of the drill string before recording begins. The sample interval is used to spread out the 1300 measurements to bracket the time that the coring tool is in the sediment. If necessary, more than one sample interval can be used, with the faster rate in the center of the expected time in the sediment and slower rates before and after, in case the center time does not coincide with the time in the sediment. However, there is no time recorded, so that the timing of data acquisition rate changes must be deduced from the programming and accurate counting of the data.

A. INTERFACE BOX

- 11 1 The temperature recorder communicates through its UART with a desktop computer interfaced through the computer's RS-232 communication port. However, the UART voltages from the temperature recorder and the RS-232 voltages from the computer are different. So an interface box was designed to change the voltage levels between the temperature recorder and the interface box. The box and its cables also provide a means to connect from the subminiature 9 pin connector on the temperature recorder to the larger 25 pin RS-232 connector on the computer.

The interface box also has a manually controlled RUN/RESET/LOAD function, which controls the program load operation from the computer to the temperature recorder. The RUN function is used to communicate with the temperature recorder. The RESET function resets the microprocessor in the recorder, and resets its program counter to 0. The LOAD function enables the recorder to load successive bytes into its memory from the RS-232 line of the desktop computer. This is how the program is loaded into the temperature recorder. When the program has run and data is to be read out, the RESET function is used to set the program counter to 0, and the data is transmitted when returned to the RUN function. This data is read by the desktop computer.

Either the PRO-350 or HP-85 computers can be selected. An HP-85 computer was used during instrument development because it was available. The HP-85 has program control of several other lines in the RS-232 connector, and these lines were programmed to control the load and reset functions that have to be operated manually with the PRO-350. Therefore, there are level changing circuits in the interface box for the reset, load, and power down/ power up lines.

The circuitry of the interface box changes the +12 volts and -12 volts signals on the RS-232 line from the computer to the +5 volts and 0 volts signals of the UART in the temperature recorder. The table below describes the signals.

SIGNAL	INTERFACE BOX		RECORDER	
	RS-232 CONNECTOR	RECORDER CONNECTOR	RECORDER CONNECTOR	
Data from desktop computer to temperature recorder	3	3	2	
Data from temperature recorder to desktop computer	2	2	1	
Signal ground	7	1	0	
+5 volts from temperature recorder	-	7	6	
+Battery	-	6	5	
Reset temp. recorder computer	5 *	8	7	
Load temp. recorder computer	8 *	9	8	
Power up (same signal as reset)	5 *	5	4	
Power down	6 *	4	3	

Table 1. RS-232 Interface Box Connections

* Connected only when the HP-85 is selected.

The data rate is 1200 baud, determined by the temperature recorder. The RS-232 data lines use negative true logic, whereas the control lines use positive true. The recorder uses positive true logic for both data and control lines.

B. DESCRIPTION OF COMPUTER PROGRAMS.

There are two types of computer programs. One type (suffix .ASC) is run by the temperature recorder, and controls the A/D converter, sample interval, storage of data, and reading data out. These programs are written in assembly language for the RCA 1802 microprocessor in the temperature recorder.

The other type of programs (suffix .BAS) are run by the DEC PRO-350, to load the 1802 programs into the temperature recorder, provide a terminal to interact with the temperature recorder's request for data collection parameters, and read the data from the temperature recorder, and convert the raw data to temperature. These programs can be run on other small computers by making suitable changes in the programs to accomodate the pecularities of the particular computer. The programs are written in BASIC for the DEC PRO-350. The PRO-350 must load the program into the recorder each time the recorder is used. The recorder has only RAM (no ROM), so the program disappears each time the power is disconnected from it. Also, part of the recording program is written over by data, when that part of the program is no longer needed, to make room for more data.

The PRO-350 programs are listed first, along with the corresponding menu item that selects them. All the programs are provided on a floppy disk, but are run from the hard disk because they need it to run during their "interim storage" phase, which is used to make room for storage of data from the recorder.

PRO 350 PROGRAMS	CORING TOOL TEMPERATURE RECORDER (MENU)
PROBASIC.BAS	
LOADER.BAS	
	 LOAD PROGRAM
LOAD1.BAS	2. LOAD PARAMETERS
LOAD2.BAS	3. DUMP DATA
LOAD3.BAS	4. DISPLAY DATA FILE
LOAD3A.BAS	5. PRINT FILE DATA
LOAD4.BAS	6. CONVERT AND PRINT FILE DATA
	7. END

PROBASIC.BAS – A short instruction program for running the MENU program.

MENU.BAS - A program to select thermal conductivity, downhole instrument operations, or exit.

LOADER.BAS - The main operating program for downhole operations. It contains the operation menu which has 7 entries (the menu is called CORING TOOL TEMPERATURE RECORDER). The load program sub-menu and program for load-ing the xxxx.ASC programs into the temperature recorder is the first item in the menu.

Menu items 2 through 6 call the sub-programs LOADI.BAS through LOAD4.BAS respectively and menu item 7 returns to the MENU.BAS program. The following sub-programs return to LOADER.BAS when finished.

LOAD1.BAS – to load operational parameters into the temperature recorder and start it running.

LOAD2.BAS - dumps data from the temperature recorder into the computer, display it to the screen, and store data to disk.

LOAD3.BAS - displays raw data for screening.

LOAD3A.BAS - prints raw data on a printer for hard copy.

LOAD4.BAS - converts raw data files to resistance and temperature files (xxxx.TXT). This program includes all the temperature recorder calibration data. The program converts the raw data, saves results to disk, and provides a hard copy on the printer.

The temperature recorder (1802) programs are loaded into the recorder by selecting item 1, LOAD PROGRAM, from the CORING TOOL TEMPERATURE RECORDER menu. The programs that can be loaded, with their corresponding menu selection and a brief description are listed below.

RECORDER PROGRAMS

LOAD PROGRAM (MENU)

MEMTS2.ASC TREC.ASC ADTEST.ASC AVERG.ASC PLYBK.ASC LOAD MEMORY TEST
 LOAD MAIN PROGRAM
 LOAD A/D TEST
 LOAD AVERAGING PROGRAM
 LOAD PLAYBACK PROGRAM

MEMTS2.ASC - Used to test memory of the temperature recorder.

TREC.ASC - The main program for data acquisition by the temperature recorder.

ADTEST.ASC - This program makes A/D readings of the temperature bridge and transmits them through the UART to the desk top computer several times a second.

AVERG.ASC - A version of TREC2.ASC which averages 8 temperature measurements by the A/D and stores the average.

PLYBK.ASC - This program is used to read out data from the recorder when the readout procedure built into TREC2.ASC fails.

The use of these programs is described in the "Temperature Recorder Operating Manual".

VI. PERFORMANCE

A. EXPERIENCE IN USING THE HPC TEMPERATURE INSTRUMENT

The first and most extensive field use of the HPC instrument to date (Jan. 1986) in the Deep Sea Drilling Project was on Leg 86 in the Western Pacific (Horai and Von Herzen, 1985). The relatively large data recovery from that leg, approximately 40 downhole measurements at six different sites, was primarily due to the dedication of K. Horai and the DSDP staff. After the first one or two sites, during which experience was being accumulated, one person (K. Horai) could maintain the instrumentation and process the data at a rate of one measurement on every other core (approximately one measurement each 3 to 4 hours). K. Horai (pers. comm., 1984) indicates that under normal operating conditions, with at least two instruments available for alternate use and preparation, one person per watch dedicated to the program could probably obtain measurements on every core (each 1-1/2 to 2 hours) during continuous operations.

Some problems were encountered on the initial use of the instrument. The difficulties with potting materials and techniques described above led to the instruments on Leg 86 being successfully potted only immediately before they were taken to sea. These potted units did not quite fit into the slots in the core nose for which they were made, so that K. Horai spent several hours on each unit to trim (sand) them down to size. We are not certain why these potted units did not fit with the same mold which was successfully used to pot trial units in the laboratory. One possibility is that inclusion of the instrument within the potting reduces the amount of shrinkage because there was less potting material to shrink compared to the trial units.

Another problem on Leg 86 was that the programming of a delay time in the instruments before taking data never appeared to perform properly. Apparently (K. Horai, pers. comm., 1984) the setting of a time delay in the program TREC (see appendix) resulted in arbitrary start times of the instrument, different than set up in the program. All of the successful runs on Leg 86 were made with the delay time set equal to zero, a condition which always started data acquisition soon after picking up the core barrel and initiation by the tilt switch. The failure with start times other than zero may be a software fault or operational confusion, as instruments tested in the laboratory appeared to perform properly.

During measurements at the first site of Leg 86 (site 576), it was noted that small oscillations of temperature over the period of HPC extension (approximately 5 to 10 minutes) were observed, apparently due to vertical heave oscillations of the ship (Fig. 9). It was observed that release of tension in the cable used to lower the HPC, as well as lowering the drill string 1-2 m, resulted in much smoother thermal decay curves (Fig. 10). Apparently the cable tension is responsible for transmitting motion to the HPC when it is extended into the bottom. Also, some data appeared to be repeated over short intervals (3-4 data points), most likely a software fault in the processing of data. This problem occurred less frequently at subsequent sites.

Other characteristics of the measurements important for obtaining accurate in-situ values were observed on Leg 86 operations as well as subsequent legs. Although, in some cases, the measurement was disturbed after the initial penetration, probably as a result of drill ship motion, sometimes a good thermal decay was observed after a few minutes of disturbance (Fig. 11). It appears that the smooth thermal decay after the disturbances can be used to obtain accurate in-situ values, using numerical methods described below. Another type of measurement is illustrated in Figure 12. Here the temperature appears to decay smoothly after the initial frictional penetration pulse, but subsequently increases irregularly thereafter. This behavior typically occurred for measurements attempted in the shallower portions of holes (<100 m), and it seems probable that the irregular temperature increases are caused by slow settling of the corer and drill string in the soft sediments. If a sufficiently long (>2 min) undisturbed record of the thermal decay exists after penetration, reasonably accurate in-situ temperature may be obtained.



Figure 9. In-situ temperature measurement on Leg 86, site 576, core 6. Note small oscillations of temperature after penetration.



Figure 10. In-situ temperature measurements on Leg 86, site 578, core 14. Note relatively smooth thermal decay after penetration. Outlined portion of record expanded in Fig. 13.



Figure 11. In-situ temperature measurements on Leg 86, site 577A, core 6. Note relatively disturbed record over about 2 min. after penetration then a smooth decay.

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The HPC temperature instrumentation was also used on several legs subsequent to Leg 86, but not as extensively. Unfortunately, three sets of instrumentation were lost on subsequent legs (87, 89 and 90), which significantly limited the amount of data obtained. According to DSDP engineers (Dave Huey, pers. comm., 1984) these losses were caused by failure of the HPC hardware, and had little to do with the operation of the instrumentation itself. Although measurements had been planned for Leg 96, none were attempted because of the unstable sediments and the fear that the leaving the HPC in the bottom without circulation, even for a few minutes, might stick the drill pipe (W. Bryant, pers. comm.). Water circulation is accomplished by pumping seawater through the drill pipe from the ship, allowing it to flow out the drill bit and up around the annulus between the drill pipe and the hole walls. This cannot be accomplished with the HPC extending out the bit. A useful development would allow seawater circulation through ports in the drill string above the drill bit after extension of the HPC. This might allow the HPC to remain extended at the same time that circulation above the HPC would prevent any cave-in or packing of sediments around the drill string.

Overall, it seems that the primary factor limiting use of the HPC is the lack of interest or motivation of shipboard scientists to obtain data. Otherwise there is little reason why such measurements could not be made an almost routine part of HPC operations.

B. EXTRAPOLATION TO EQUILIBRIUM TEMPERATURES

During the time the HPC is in the bottom, the frictional heat generated by the penetration is conducted away to the surrounding sediments (Fig. 10). The two-dimensional radial conduction theory for the physical configuration of the HPC has been solved numerically by K. Horai (pers. comm., 1984). Both the theory and experimental field data show that the excess temperature of the HPC over in-situ values is reduced to about 15-20% of its initial value after about 10 minutes in the sediment. It is not practical to leave the extended core barrel in the sediment much longer than this, so that an accurate extrapolation of the frictional heating pulse is necessary.

A comparison between the theory of Horai and an actual measurement is shown in Figure 13. The agreement between theory and practice is extremely close after the first few data points (40-60 sec.). The lack of agreement during the first few data points occurs when the temperature distribution in the HPC wall is probably not uniform. A small remaining problem seems to be that the theoretical temperatures fall slightly below the temperatures measured near the end of the thermal decay. This may be due to the fact that the anomalous heat in the HPC is being conducted axially, as well as radially. The excess heat in the HPC is conducted preferentially toward the thin end of the wedge shaped cutting shoe, where the thermistor is located (Fig. 6). This would tend to cool the HPC more rapidly than if the HPC were infinitely extended along its axis, as assumed in the two-dimensional theory.

Therefore, in the measurements for Leg 86, Horai and Von Herzen (1985) assume that the true in-situ equilibrium temperature lies between the last measured point during the HPC penetration, and the theoretical



Figure 13. Thermal decay of HPC_during core 14, hole 578, Leg 86 (see Fig. 9). Solid dots are data points fit on the theoretical curve shown.

extrapolated value. Probably the true in-situ temperature is much closer to the value extrapolated using the two-dimensional theory, as seems to be the case from the comparison of the curves (Fig. 13). This should be tested by attempting to leave the HPC in the bottom as long as possible so that the data may be compared with the theoretical extrapolation.

Another useful set of data is the temperature of the near-bottom water measured by the HPC during its passage through the drill pipe. The temperature of the water in the drill pipe is usually close to that of the surrounding sea water, so that a calibration of the instrumentation can be obtained for each measurement. This is particularly useful if different sets of instrumentation are used for measurements in the same hole. It provides a useful check on the accuracy of the data obtained, and also provides another temperature data point at the sea floor.

C. TEMPERATURE CALIBRATION STABILITY.

We have some limited data on the temperature stability of some units, based on accurate repeated calibrations. Altogether 10 hybrid electronic units were constructed. One was not calibrated, because it was left as a hybrid that could be adapted to fit the annular cavity of the new cutting shoe, with the possible addition of acoustic communication. Three others were lost during operations at sea shortly after they were constructed and calibrated; hence, no history of stability over time exists for these units. Of the six other units, four (units Nos. 5, 6, 7, and 8) were calibrated in the laboratory two or more times, which provides evidence of temperature stability.

Calibrations were performed with the entire unit immersed in a stable water bath, at 7 temperatures uniformily distributed at about 10°C intervals over the range 0 to 60°C, as described above. The differences between calibrations in 1983, over a period of about 11 months for units Nos. 5 and 6, and about 7 months for units Nos. 7 and 8, are shown in Fig. 14a. Similarly, the differences over about 11 months in 1984 for all 4 units are illustrated in Fig. 14b.

The equivalent temperature shifts shown by the 1984 calibrations are within $.01 - .02^{\circ}C$ over the entire temperature range for all units, and for 2 of the units (Nos. 7, 8) in the 1983 calibrations. These are probably within the resolution of measurement for these units. The larger shifts $(.05 - .07^{\circ}C)$ for units Nos. 5 and 6 in 1983 may correspond with their use on Leg 92 and subsequent legs. The shifts are fairly uniform over the entire temperature range, suggesting that the thermistor sensor (most likely) or a bridge resistor may have changed as a result of physical shocks associated with the measurements. (Note that units Nos. 5 and 6 shifted in the opposite sense from each other.) Unfortunately, the units used extensively on Leg 86 were lost in operations on subsequent legs before they could be recalibrated.



Figure 14. Comparison of temperature calibrations of 4 instruments before and after deployment and use at sea, over dates as indicated. Differences are calculated as the later calibration subtracted from the earlier.

VII. <u>FUTURE</u>

The current software for the temperature recorder provides a sampling interval in the range of 0.1 to 6553.5 sec. The 0.1 sec sampling interval corresponds to an upper frequency limit of 5 Hz (the Nyquist sampling rate) or 2 Hz (if you want to preserve the waveform). This rate would fill the memory in 130 seconds. The longest sampling interval of 6553.5 sec corresponds to a minimum period of 1.82 hr (at the Nyquist rate), or 4.55 hr (for waveform preservation). This slowest rate would take 2366 hr to fill the memory, which far exceeds the battery life (approximately 20 hrs.).

Other applications are possible for the temperature recorder. The input is an A/D converter, so it could be adapted to measure any voltage or resistance bridge measurement. A possible application is to measure the output of a strain gage type pressure transducer, or the voltage output of any sensor that has a voltage output. The A/D converter can sample up to 25 times per second, if the integrating capacitor is changed and the software modified. This implies an upper frequency limit of 12 Hz (the Nyquist rate) or 5 Hz (waveform preservation). The lowest frequency is limited by the memory size and the sampling rate. At 25 samples per second, the memory would be filled in 52 seconds. At the other extreme, data rates may be as long as desired to fill the memory (currently 1300 2-byte words), limited only by battery life.

The present coring shoe cavity has been enlarged in a new design, which has the same diameter as the original cavity, but it extends completely around the coring tool to form an annulus. It is 0.28 in. thick and 5 in. long. This larger cavity, if layed out flat, would be 10 in. long, by 5 in. wide, by 0.28 in. thick. It was possible because the strength of a shell (the cylinder) is much stronger than a flat plate. Actually, the small cavity is probably stronger that we calculated (see Appendix A), but we conservatively used the formula for a flat plate. This new cavity was made by welding the outer wall to the piece containing the cutting edge and the inner wall. It was no more expensive than the previous design, because no EDM (electro discharge machining) was necessary. It was designed at the DSDP at Scripps. There is some question about the strength of the cylinder wall under impact, but that can be tested. If it would fail under severe impact, you could take your chances, and accept a loss if it does fail. However you would not want it to be damaged, then deform under pressure and get stuck in the drill string.

The whole diameter of the coring tool can be used for instrumentation where only drilling and logging is being done and the material being drilled is too hard for HPC coring. This increases the space available for additional instrumentation.

Each time data is read from the temperature recorder, the coring shoe has to be opened up to enable connection to the instrument, and cleaned and closed up for the next use. This is troublesome, as well as risky by exposing the electronics to possible damage from salt water, or the chance of losing data by inadvertently disconnecting the battery. The temperature recorder could use acoustic communication through the wall of the cavity in the coring shoe. A small transducer about the size of a dime could transmit using a 1 MHz carrier, using frequency shift keying. It could also receive instructions. With the large annular cavity, there would be plenty of room for batteries, certainly enough to last one leg (2 months), or possibly even for a year. So the instrument would not have to be opened up during use, or possibly not at sea. It could be sent back to the lab for new batteries.

At one time flexible printed circuits were considered as an alternative to the hybrid. A flexible circuit was tried in a rolled up circuit for another instrument, and it was discovered that the traces break where they are soldered to the integrated circuit packages. The flexible board industry laminates rigid boards to the flexible circuit in the regions where there are electronic parts, and uses the flexible part only to interconnect the rigid boards. If such a structure were used in the hollow annulus, it would take up most of the space. The resistance to physical shock is problematical. However, the hybrid is difficult and somewhat expensive to produce in small quantities, which might be alleviated with a more spacious layout using small outline package IC's on boards with flexible interconnections. Another possibility is combining multiple hybrids on separate hard boards, with flexible interconnections.

There has been a desire expressed to reach higher temperatures, perhaps 200°C. The present design has a nominal limitation of 75°C, determined by the temperature rating of the A/D converter. All other parts are rated to 125°C. The batteries can last a short time at 70°C. Eutectic solder (63% tin, 37% lead) melts at 183°C, and other ratios get mushy at this temperature.

To reach 200°C, the entire electronics circuitry and power supply would have to be redesigned with higher temperature parts. Such high temperatures would normally be encountered at depths or in regions (e.g., ridge crests) where the materials are too hard to use the HPC. If the entire diameter of the coring tool could be used for instrumentation, then existing higher temperature batteries (0.55 in. diameter) could be considered. Hybrid circuits at 200°C could probably be fitted into the 0.28 in thick space inside the wall of a coring tool. The thermometer would be a platinim resistance thermometer of some type of rugged construction. Absolute accuracy of 0.5°C seems possible at the higher temperatures.

VIII. ACKNOWLEDGEMENTS

Several persons have contributed significantly to this project. Ed Mellinger wrote the assembly language programs for the recorder and participated in the design of the electronics and construction of the breadboard. Skip Pelletier wrote the programs for the personal computers (Apple and Pro-350) for controlling the temperature recorder, and processing the recorder data. Dave Huey and Mike Storms (DSDP) designed and arranged for construction of the special coring shoe. Don Bellows (DSDP) did the same for the new annular cavity coring shoe. Charlie Yiakas and Dave Vacca (TSI) were always helpful and patient in consultation and for construction of the hybrid circuit. Ki-iti Horai gave the recorder its first extensive field tests on Leg 86 of the DSDP, and developed the theory to obtain equilibrium temperatures from transient thermal decay of the corer.

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APPENDIX A. CAVITY STRENGTH CALCULATIONS

1. Kidney shaped <u>cavity (currently used)</u>.

The cavity had to fit into the 9/16 in. thick wall of the coring shoe, and operate at 21,000 ft. deep (10 kpsi) with some margin of safety. A very strong heat treatable stainless steel, Nimark 250, from Carpenter Technology, Carpenter Steel Division, Reading, PA, with a typical yield strength of 250 kpsi, was used for the portion of the coring shoe containing the cavity.

We selected a hybrid package to fit in the cavity. We made the width just large enough to accomodate the hybrid package (0.80 in.) and then made the cavity as high as possible. The actual cavity walls are curved (Fig. 6), so although the height of the cavity is 0.28 in., the height of a rectangular hybrid package inscribed in the curved cavity is less. The length of the cavity is 5 in. to accomodate the combined package of hybrid and printed circuit board. A cylindrical hole 0.125 in. in diameter extends beyond the end of the cavity to accomodate the thermistor probe.

A flat plate approximation was used for the curved walls of the cavity. This approximation is conservative, in that the curved shape should be stronger. The width of the cavity used for the flat plate approximation is the portion of the arc of the outer wall that has uniform wall thickness, measured along the outside surface of the coring tool. The equation below is derived from formulas for flat plates with straight boundaries and constant thickness (Roark and Young, 1975, p. 386, Table 26).

The rectangular plate, shown in Figure A-1, has three edges fixed and one edge simply supported. The force is uniform over the entire plate. Using Table 26, Case 9 (Roark and Young, 1975, p. 394), the general form for the stress is:

 $s = \frac{B q b^2}{t^2}$ (A.1)

where a, b = the dimensions of the plate; t = thickness of the plate; q = unit force on the plate; and B is a variable which depends on the ratio a/b and the direction and location of the stress. B is taken from a table in the above reference which had to be extended to a/b = 0.16 (vs. a minimum of 0125 in the table) by graphical extrapolation on log-log graph paper.



Figure A-1. Flat plate model used for cavity strength calculations.

For q = 10,000 psi, a = 0.800, b = 5.00, t = 0.145, the maximum stress in the a direction (circumferentially) occurs at $x = \pm a/2$, z = 0.6 b, for which B = 0.0133,

The maximum stress in the b direction (axially) occurs at x = 0, z = 0, for which B = 0.0083,

$$(0.0083)(10^4)(5)^2$$

 $s_b = ----- = 98,690 \text{ psi}$
 $(0.145)^2$

The largest stress is 158 kpsi in the outer wall, which is about 70% of the minimum yield strength (230 kpsi) of the material. The curvature of the plate serves to strengthen the plate further, but is not included in the calculation. The coring shoe was tested at and survived 10 kpsi, the working pressure. It was not tested at a higher pressure for fear of collapsing the cavity. However, we infer from the calculations for the . strength of an annular cavity (see following) that the actual strength of this cavity design may be considerably stronger.

The cavity for the battery was made the same size as the electronics, except for length. It is 3 in. long, while the cavity for the electronics is 5 in. long.

An annular cavity, the same thickness as the electronics cavity and 0.25 in. deep, provides space for compression of the air when the two parts of the coring shoe are screwed together. Without this space, the air compressed in the cavity would bend the top of the lid of the hybrid package and short out the hybrid circuitry.

2. Annular Cavity.

A much larger space for the electronics can be achieved if an annulus were hollowed out of the interior of the coring shoe. A cylinder is stronger than a flat plate. A short cylinder is stronger than a long cylinder because the support from the ends has more influence on a short cylinder. The cylinder can fail either in collapse or in buckling. Different equations are used for each case, and the lowest pressure at which the cylinder fails is the one that is used. A number of lobes are formed by the tube in buckling. The buckling pressure for each lobe number is calculated and the lowest buckling pressure is the one used.

The dimensions of the original coring shoe (Fig. 6) were used, which would make the same thickness available for the electronics. First the collapse equations are used, which result in a collapse pressure of 25,350 psi for the inner cylinder and 17,925 psi for the outer cylinder. Then the pressure for buckling the cylinder for a given number of lobes is calculated versus the cylinder length. Plots were made of pressure vs. length for each number of lobes. Then, using the minimum buckling pressure of all the curves, length or pressure can be chosen. We chose 5 in. long, which gave a buckling pressure of 18,000 psi, with 3 lobes.

Cylinder collapse strength.

The walls of the annulus are concentric cylinders, each considered as a short supported tube (Fig. A-2). The cylinders are fixed at one end and simply supported at the other.

The material used is Nimark 250, with the following characteristics:

> Min. yield strength = 230 kpounds/in² Modulus of elasticity, E = 27.5x 10⁶ Poissons ratio, v = 0.3

The equations below are from Table 32 (Roark and Young, p. 504): Formulas for a thick walled vessel under internal and external loading. Notation: a = outer radius; b = the inner radius; $s_2 = normal stress in$ the circumferential direction; v = Poissons ratio.

The inner cylinder has uniform internal pressure, q lb/in², in all directions; ends capped. Using Table 32, Case 1b (Roark and Young, p 504),

 $\max s_2 = q - \frac{a^2 + b^2}{-----}$ at r = b $a^2 - b^2$

For

a = 1.365b = 1.222

 $(1.365)^2 + (1.222)^2$ max $s_2 = (10^4)$ = 25,350 psi (1.365)² - (1.222)²



$$\max s_{2} = \frac{2 q a^{2}}{a^{2} + b^{2}} \quad \text{at } r = b \quad (A.3)$$





(A.2)

Elastic instability calculations.

The outer cylinder is subject to elastic instability because it has external pressure. A cylinder with supported ends has a higher pressure for elastic instability, as it becomes shorter.

The equation below is from Table 35, Formulas for elastic stability of plates and shells (Roark and Young, p. 556). Notation: E = modulus of elasticity; v = Poisson's ratio; and t = thickness for plates and shells.

The thin tube has closed ends under uniform external pressure, lateral and longitudinal (length of tube = ℓ ; radius of tube = r). The ends are held circular. Using Table 35, Case 20 (Roark and Young, p. 556), where q' = external pressure at which elastic buckling occurs:

$$q' = \frac{E\left(\frac{t}{r}\right)}{1 + \frac{1}{2}\left(\frac{\pi r}{n\ell}\right)^{2}} \left\{ \frac{1}{n^{2}\left[1 + \left(\frac{n\ell}{\pi r}\right)^{2}\right]^{2}} + \frac{n^{2}t^{2}}{12r^{2}(1-v^{2})}\left[1 + \left(\frac{\pi r}{n\ell}\right)^{2}\right]^{2} \right\}$$

here n = number of lobes formed by the tube in buckling. As described by Roark and Young, to determine q' for tubes of a given t/r, a group of curves is plotted for each integral value of n of 2 or more, with ℓ/r as ordinate and q' as abscissa; that curve of the group which gives the least value of q' is then used to find the q' corresponding to a given ℓ/r .

Plots of this equation are shown in Fig. A-3, where pressure is plotted vs. length of the cylinder, for r = 1.785 in. and t = 0.145 in.

The buckling pressure is 18 kpsi, for a length of 5 in., which is nearly the same as the collapse pressure for this shape. Thus, the walls of the 5 in. long cylindrical annulus are more than strong enough to withstand 12 kpsi at the bottom of the drill string.



Figure A-3. Buckling pressure vs. length of cylinder, for various numbers of buckling lobes.

APPENDIX B. DETAILED DESCRIPTION OF ELECTRONICS

1. <u>Description of Block Diagram</u>.

The block diagram (Fig. 4) corresponds very closely in structure to the actual schematic. The temperature recorder consists of 7 blocks plus battery: the microprocessor, RAM, UART, A/D converter, thermistor bridge, power supply, and battery. The measurement begins with the thermistor bridge, with a voltage nearly linear with temperature. The A/D converter digitizes this voltage with 13 bit resolution. The microprocessor instructs the A/D converter when to convert and stores the measurement in memory. The UART is used to send the stored data serially to an awaiting external device (a computer). The UART is also used to load the program into RAM for 2-way communication with the computer.

A voltage regulator provides 5 volts to the electronics from an 8 volts battery. A 4 volts mode is available to reduce power consumption. The power switch controls the 5 volts to the A/D converter, so that it may be shut down when not being used, in order to save power. The voltage inverter converts some of the 5 volts to -5 volts for the A/D converter.

2. Detailed Description of the Schematic (Fig. B-1).

The 1802 microprocessor controls all the rest of the elements of the temperature recorder. A crystal of 38.4 kHz runs with the internal oscillator of the microprocessor to provide the computer clock. A low frequency was chosen to save power, since the microprocessor does not have to work very fast. The clock frequency was chosen so that a binary divider chain can provide the UART clock input, which is 16x the UART baud rate.

There are 3 kbytes of RAM, controlled thru the 1866 memory selector, by the address lines of the microprocessor. The 8 address lines are multiplexed. The high order byte comes out first and the low 4 bits are stored in the 1866. Two bits drive the RAMs directly and two are decoded to select the RAMs. The low order byte comes out second and drives the address lines of the RAMs directly. The RAMs are 1 k x 4 bits, so 2 RAMs are used simultaneously: one for the upper four bits of data, and one for the lower 4 bits of data. A write enable line from the computer is active when writing to memory. The address lines and the inhibit for the 1866 selector come outside the hybrid so it can be tested, more memory added or other memory substituted for the internal memory.

The UART runs at 1200 baud, and provides 2-way communication thru the 9 pin recorder connector. The clock for the UART is 19.2 kHz, which is 16x the 1200 baud rate, derived from a binary divider on the computer clock. The baud rate can be changed on the printed circuit board.

The UART 8 bit data busses for transmitting and receiving are connected to the microprocessor data bus. The transmit buffer is loaded by I/O line 5, and the receiver buffer is read by I/O line 4. These lines come from the 1853 N line decoder. When data has been transmitted, the transmit buffer



Figure B-1. Schematic of the temperature recorder.

empty line raises (to inverted logic O) the external flag 1 input (EF1-bar) of the microprocessor. When the receive buffer is full, the data ready line raises the external flag 2 input (EF2-bar) to the microprocessor. The microprocessor has to test these flags to determine the state of the UART.

A dual slope type A/D converter is used to measure the bridge output. Its integration period is set to 1/60 second to average out 60 Hz pick-up while making lab measurements. The A/D converter averages out noise over the integration period, which means it is less sensitive to noise.

The run/hold line for the A/D is controlled by I/O line 3. This line, as well as the high byte enable and low byte enable for data output are driven by buffers that get their power from the A/D power, but do not load the driving source when the A/D power is switched off.

The reading of the high byte out of the A/D is enabled by I/O line 1, and the low byte by I/O line 2. A tri-state buffer is used between the A/D data output and the data bus so the A/D converter can be turned off without pulling the data bus down.

The bridge circuit for the thermistor is made of Vishay (Malvern, PA) resistors, which are very stable and temperature insensitive (<lppm/°C). The bridge runs at a low voltage so that only about 10 m°C self heating of the thermistor occurs. The self heating, which varies about 30% over the temperature range, is included in the temperature calibration. A zener-like voltage regulator provides 2.5 volts to the bridge to keep the small fluctuations (25 mV) in the 5 volts line from changing the voltage to the bridge during conversion.

The inactive leg of the bridge also provides the reference voltage for the A/D converter, which is set to about 200 mV. A ratiometric measurement is made from the bridge voltage, so the exact voltages are not important. A convenient feature of this A/D converter is the Input Lo terminal, a floating high impedance input, which allows us to center its operating range in the middle of the bridge output swing. Thus we can use both the positive and negative range of the 12 bit plus sign A/D converter to obtain a 13 bit conversion.

An I/O instruction has numbers 1 thru 7, which is encoded into 3 bits on the NO, N1, and N2 lines. The 1853 N line decoder decodes these 3 bits into 8 lines for operating various I/O devices. These are:

- 1. IN 1 high byte enable on A/D.
- 2. IN 2 low byte enable on A/D.
- 3. OUT 3 Run/Hold on A/D.
- 4. IN 4 read UART receive buffer.
- 5. OUT 5 load UART transmit buffer.
- OUT 6 raise power supply voltage to 5 volts (also available as an output pin for control of an external device).
- 7. OUT 7 lower power supply voltage to 4 volts.

Although these lines can be programmed to be inputs or outputs, they are wired to do only one or the other (except for the potential use of I/O line 6).

A very low power voltage regulator supplies a regulated 5 volts to the electronics, when the computer is running normally. It can be switched by I/O line 7 to regulate at 4 volts for standby to hold the data at about 1/2 the current (about 0.5 mamp) as at 5 volts. Under program control the voltage is switched to 4 volts after the data has been acquired. The computer seems to run at 4 volts, but proper operation of the computer and memory is not guaranteed below 4.5 volts. The power can be raised to 5 volts with I/O line 6.

To further save power after the data has been acquired, an Idle instruction stops the microprocessor, while the clock continues to run. This chops the current by another factor of 2 to about 0.25 mamp when the supply voltage is 4 volts. As it is wired up (no interrupts), the microprocessor cannot get out of the idle condition by itself, but must be reset externally.

A power switch, controlled by the Q line of the microprocessor, switches the +5 volts for the A/D converter. A 7660 voltage inverter, powered by this switched +5 volts, generates the -5 volts for the A/D converter. The A/D converter consumes about 2.5 mamp when on, so considerable power is saved when it is turned off.

The 1802 has a convenient Load mode, using the DMA IN input. Data from the UART is read into the memory, starting at address $0000_{\rm H}$, when in load mode and the DMA IN line is low. The Data Ready line of the UART controls the DMA IN line (via a gate) when in the Load mode. The load mode is entered by raising the reset and the load line in that order.

APPENDIX C. DETAILED LIST OF INPUT/OUTPUT INSTRUCTIONS

Details on the Temperature Recorder Computer Instructions.

Below are instructions specific to this particular computer. See the RCA 1802 programming manual for the general programming of the 1802.

The memory is 3 kbytes (3 x 1024), which is addressed sequentially from $0000_{\rm H}$ through $0BFF_{\rm H}$ (3071_o). The same memory is addressed each 4 kbytes up to 64 kbytes, because the address decoder does not look at the upper 4 bits.

Input/Output Instructions

IN 1	read A/D high byte
IN 2	read A/D low byte
OUT 3	run/hold-bar A/D converter (it takes 1/15 sec to make a conversion)
IN 4 OUT 5	read UART receive buffer load UART transmit buffer
OUT 6 OUT 7	raise power supply voltage to 5 volts lower power supply voltage to 4 volts

Other Instructions

SET Q switch power on to A/D RESET Q switch power off to A/D

FLAGS (note: actual flag inputs are inverted logic)

- EF1 UART transmit buffer empty
- EF2 UART receive data buffer full

EF3 A/D status: 1 = running; 0 = finished, or ready to start

EF4 tilt switch vertical (start program)

Load Memory with Program

Raise reset line to 5 volts, then raise load line to 5 volts.

Now any input to the UART will go into memory. All 8 bits from the UART input go into memory.

To terminate load, lower the load line to 0 volts, then lower the reset line to 0 volts.

Now the program will start at location 0000.

UART

The UART runs at 1200 baud, 8 bits, no parity. 1 = 5 volts, 0 = 0 volts.

APPENDIX D. BRIDGE CIRCUIT EQUATIONS

The bridge circuit for the thermistor is shown in Fig. D-1.

 E_1 is the input voltage to the bridge.

 E_{1N} is the voltage into the A/D converter. The negative input to the A/D converter is derived from the inactive leg of the bridge, and is at the center of the bridge excursion.

 E_{REF} is the reference voltage into the A/D converter. The reference voltage is 1/2 the full scale input to the A/D converter. This reference voltage is derived from the inactive leg of the bridge, instead of from a E_1 separate divider.

 R_3

R4

 R_5

ref

The A/D converter is ratiometric, that is, measures E_{1N}/E_{REF} . Therefore, the value of E_1 is not important, as long as it is stable shortly before and during a measurement.

The circuit can easily be solved for R_2 , the thermistor resistance, as follows:



Figure D-1. Thermistor bridge circuit.

R₁

 R_2

(Thermistor)

in

or
$$R_2 = 8.715 \begin{bmatrix} (3.734039 + count) \\ 2048 \end{bmatrix} \begin{bmatrix} (3.023975 - count) \\ 2048 \end{bmatrix}$$

with the following circuit values (Fig. D-1):

 $R_{i} = 8.715 \text{ K} \\ R_{2} = \text{thermistor resistance} \\ R_{3} = 2.84671 \text{ K} \\ R_{4} = 2.57377 \text{ K} \\ R_{5} = 0.94135 \text{ K} \\ \frac{E_{1n}}{E_{ref}} = \frac{\text{count from A/D}}{2048}$

APPENDIX E. CURVE FITTING TO THERMISTOR CALIBRATIONS

We have found that the thermistor calibrations are fit very well (to better than a few millidegrees over a 60° range) by a 3-parameter equation (Steinhart and Hart, 1968)

$$T^{-1} = A + B \ln R + C(\ln R)^3$$

where T = temperature (K) R = thermistor resistance (ohms)

A, B, C are constants to be determined for each thermistor. Obviously, at least 3 calibration values are needed to determine the 3 constants in the equation, but usually we have an overdetermined problem of finding the best solution for more than 3 values. This solution is found by the usual least-squares method. Notice that the squared term of the power series is absent, so the usual curve fitting routines for a power series are not used. The least-squares method for this equation, assuming equal weighting of calibration values, is derived as follows:

We simplify the notation by solving for the equation

 $X = A + By + Cy^3$

where we have substituted x for T^{-1} , and y for ln R. Then we want

to determine the constants A, B, and C such that $\sum (A + By + Cy^3 - x)^2 = minimum$ for n calibration values. This condition is met by differentiating this expression with respect to each of the constants and setting equal to zero:

 $\frac{\partial}{\partial A} = 2 \sum (A + By + Cy^3 - x) = 0$ $\frac{\partial}{\partial B} = 2 \sum (A + By + Cy^3 - x) y = 0$ $\frac{\partial}{\partial B} = 2 \sum (A + By + Cy^3 - x) y^3 = 0$ $\frac{\partial}{\partial C} = 2 \sum (A + By + Cy^3 - x) y^3 = 0$

where the summation over n is assumed

or:

$$\sum (A + By + Cy^{3} - x) = 0$$

$$\sum (Ay + By^{2} + Cy^{4} - yx) = 0$$

$$\sum (Ay^{3} + By^{4} + Cy^{6} - y^{3}x) = 0$$

Re-arranging terms:

An + B
$$\sum y$$
 + C $\sum y^3 - \sum x = 0$
A $\sum y$ + B $\sum y^2$ + C $\sum y^4 - \sum yx = 0$
A $\sum y^3$ + B $\sum y^4$ + C $\sum y^6 - \sum y^3x = 0$

Rewriting in matrix format:

$$\begin{bmatrix} n & \sum y & \sum y^{3} \\ \sum y & \sum y^{2} & \sum y^{4} \\ \sum y^{3} & \sum y^{4} & \sum y^{6} \end{bmatrix} \qquad . \qquad \begin{bmatrix} A \\ B \\ C \end{bmatrix} = \begin{bmatrix} \sum x \\ \sum yx \\ \sum y^{3}x \end{bmatrix}$$

or:

i.

$$\left[\mathsf{D}_{i,j}\right] \cdot \left[\mathsf{B}_{i}\right] = \left[\mathsf{E}_{i}\right]$$

The desired solution is therefore:

 $\begin{bmatrix} B_{i} \end{bmatrix}^{*} = \begin{bmatrix} D_{i} \end{bmatrix}^{-1} \quad . \quad \begin{bmatrix} E_{i} \end{bmatrix}$

APPENDIX F. PROGRAM LISTINGS FOR THE TEMPERATURE RECORDER 1. INSTRUCTIONS FOR TREC

NAME: TREC/AVRG

PURPOSE: To control operation of DSDP Coring Tool Temperature Recorder

MACHINE: 1802 COSMAC

SOURCE LANGUAGE: RCA CSDP Assembler

DESCRIPTION:

The Coring Tool Temperature Recorder is a miniature data aquisition system packaged in a custom hybrid integrated circuit. It was developed at WHOI for the purpose of measuring in situ temperature of deep ocean sediments, from the cutting shoe (tip) of a Hydraulic Piston Coring (HPC) tool. The recorder contains an 1802 microprocessor, 3k bytes of static RAM, a serial I/O port, and an A/D converter on a substrate approximately 0.55" x 2.7". An external sensor (nominally a thermistor bridge) and battery pack complete the recorder, which has the capability to aquire 1200 to 1500 fourteen bit data points at intervals of 0.2 seconds to several hours.

TREC is an operating program designed to drive this recorder. During initialization it allows the user to interactively specify one to eight sets of data samples to be aquired by the recorder, with independent turn on delay, sample interval, and sample count for each sample set. After initialization the sample parameters are echoed back to the user for verification, and the recorder's program memory is checksummed. Upon user command TREC enters aquisition mode, and each sample set is aquired using the specified delay, sample count, and sample interval parameters. TREC terminates aquisition and places the recorder in a low power idle mode after all specified sample sets have been aquired, or when the available data memory becomes full. After recovery of the recorder, the recorder CPU is reset and TREC outputs the stored data values to an external data storage device.

OPERATION:

LOADING

1. Place the recorder in Load mode and load the TREC binary code file into the recorder from the support computer. Detailed instructions for this proceedure will be provided for each type of support computer which is used.

2. Reset the recorder CPU to begin execution of TREC. Again, instructions will vary depending upon the support computer provided.

INITIALIZATION

3. TREC will prompt the user for three parameters for each sample set: the number of samples, the sample interval, and the delay period prior to sampling. Each parameter is input as a 0-4 digit <u>hex</u> number followed by a carriage return, as further specified in the "Input" section below. These <u>sample parameters</u> are stored in the recorder memory and later used to control operation of the recorder during data aquisition. The sample parameter prompts are: # SAMPLES ? Input the desired hex number of samples for this sample set. The sum of samples for all sample sets should be less than 1380 decimal (564 hex). Entering a value of zero samples terminates the initialization proceedure and causes verification to begin.

INTERVAL ? Input the desired hex sample interval for this sample set, in tenths of seconds, from a minimum of 2 up to a maximum of 65535 decimal (2-FFFF hex). Note: a specified interval of 1 (0.1 sec) will result in an actual value of 65536 (6553.6 sec).

DELAY ? Input the desired hex value of turn on delay prior to this sample set, in seconds, from 0 to 65536. Turn on delay is the delay before aquiring the samples specified by # SAMPLES and INTERVAL.

VERIFICATION:

When a value of zero samples is input during initialization, input stops and verification begins. Each of the sets of sample parameters is read from memory and output for inspection by the user. A simple checksum is computed over program memory and output also. The user is then prompted for further action. A typical verification output would be:

#SAMPLES =	0080	INTERVAL =	0064	DELAY =	1CO2
#SAMPLES =	0400	INTERVAL =	000A	DELAY =	0000
#SAMPLES =	00E4	INTERVAL =	0032	DELAY =	0000

CHECKSUM = 0058 ACTION ?

(This would specify an inital delay of 120 minutes followed by 128 samples at 10 second intervals, 1024 samples at 1 second intervals, and 228 samples at 5 second intervals, followed by power down.)

Examine each displayed parameter to be sure it is correct, and compare the checksum to the correct value for the TREC version in use. If all information is correct, typing a capital "G" in response to the "ACTION ?" prompt will cause TREC to enter <u>aquisition</u> mode and begin the first delay period. Typing "G" may be delayed for as long as necessary, to synchronize the pending aquisition cycle with external events such as HPC operation.

If any of the displayed sample parameters are incorrect or must be changed, TREC may be re-initialized by typing a carriage return in response to the "ACTION" prompt, and returning to step 3 above.

If the checksum does not match the correct value given for the TREC version in use, a data error has occurred during or after program load. Do not use a program with a checksum error. Reload TREC from the support computer and start over.

AQUISITION:

During the delay period prior to sampling, TREC outputs an "@" symbol every second to verify correct program operation. Once the "@" is observed the instrument may be sealed and deployed using appropriate proceedures. The <u>tilt switch</u> is active during the delay periods, as follows: If the recorder is horizontal continuously for more than 1 minute, the delay timing stops ("@" stops also). Delay timing restarts from the initially specified delay value for the current sample set, after the recorder has been restored to vertical continuously for 5 minutes.

After the specified delay period has elapsed the A/D converter is powered up and the number of samples specified for the current sample set is aquired, at the interval specified. After sampling, the number of samples in the <u>next</u> sample set is read from memory. If it is zero, the recorder is powered down and the CPU executes an "Idle" instruction until instrument recovery. If the number is not zero, the remaining sample parameters for the next sample set are read from memory and the next delay period begins. The A/D converter is powered down during delay unless the delay is zero seconds. If data memory becomes full during sampling further sampling is inhibited and the recorder is powered down as usual.

PLAYBACK:

Upon recovery the recorder will normally have exited Aquisition mode and be powered down with the CPU reset. After power up and release from reset TREC will dump the stored data as a sequence of 4 hex digit records, one for each sample aquired. When all samples have been dumped the message "STOP" is output and the recorder is powered down. The data playback sequence may be restarted after power down by again resetting the recorder, using the playback routine on the support computer.

INPUT FORMATS:

All input is serial asynchronous ASCII at 1200 baud; 8 data bits, one stop bit, no parity bit.

SAMPLE PARAMETERS. Each parameter is input as a zero to four digit hex number followed by a carriage return. Valid hex digits are echoed back to the user input device; invalid or unrecognized characters are not. The carriage return is not echoed. Valid digits are 0-9 and A-F. If no digits are input before the carriage return a default value of zero is assumed for the parameter. Only the last 4 digits entered are significant, so that typing errors may be corrected by typing unitl the last 4 digits represent the desired value and then typing carriage return.

OUTPUT FORMAT:

All output is serial asynchronous ASCII at 1200 baud; 8 data bits, one stop bit, no parity bit.

PROMPTS. The prompt "#SAMPLES?" is output upon initial reset, followed by the "INTERVAL?" and "DELAY?" prompts. The set of three prompts is repeated for each sample set specified. The prompt formats are:

(CR)(LF)#SAMPLES?(SP)(ETX)
(SP)(SP)(SP)INTERVAL?(SP)(ETX)
(SP)(SP)(SP)DELAY?(SP)(ETX)

where LF = line feed (hex OA), CR = carriage return (hex OD), SP = space (hex 20), and ETX = End-Of-Text (hex O3). A 200 millisecond pause follows each CR transmitted by TREC.

VERIFICATION. A double linefeed is output on entry to the verification routine. A verification record is output for each set of sample parameters entered during initialization. Each record has the format

#SAMPLES= HHHH INTERVAL= HHHH DELAY= HHHH(CR)(LF)

where HHHH is a four digit hex number. After the last record the checksum is output in the format

CHECKSUM= HHHH ACTION?(SP)(ETX).

AQUISITION. During the delay intervals the character "@" is output once per second. This output stops during tilt switch waits and during the actual sampling process.

PLAYBACK. After recovery and CPU reset the recorded data is output. Each 14 bit sample is output as four hex digits H_i in the following format

 $H_3H_2H_1H_0(LF)(CR)$

where: H₃ contains the overrange (bit 2) and polarity (bit 3) bits; bits 0 and 1 of this digit should be ignored. H₂ contains bits 11 (MSB) through 8 of the A/D output, H₁ contains bits 7 through 4, H₀ contains bits 3 through 0 (LSB).

Each carriage return is followed by a 200 millisecond pause. After the last aquired sample has been output, the end of data message is output:

STOP(LF)(CR).

ERROR MESSAGES:

• .;

TREC performs minimal input validity checking due to its small size. If a non hex character is supplied during data input, that character will not be echoed to the user. No error message is output. Input values are not checked for reasonableness or consistency.

RESTRICTIONS:

1. The sum of the number of samples in all sample sets should not exceed 1380 decimal. No harm will result from specifying a greater number of samples, but only the first 1380 samples will be aquired.

2. Specifying a sample interval of "1" (0.1 sec) will result in an actual sample interval of 6553.6 seconds.

SUBROUTINES REQUIRED:

(all internal to TSEC package)

The first set of routines listed below are memory resident before, during, and after aquisition.

PLAYBK	-	dumps recorded samples.
OUTDAT	-	outputs a 4 hex digit number.
MAIN	-	aquisition control program.
DLAY	-	turn on delay; tilt switch.
AQUI	-	sample and sample interval control.
GETDAT	-	gets one sample from A/D.
TSEC		precision 0.1 second timing routine.
OUTCHR	-	outputs a variable length ASCII string.

The following routines are loaded into the data area of memory and are executed once during initialization. They are overlaid (destroyed) by data during the actual aquisition process.

INIT	-	initializes registers, prompts for parameters.
VERIFY		outputs parameters, computes checksum.
INDAT	-	reads and converts 4 digit ASCII hex numbers.

DATA STRUCTURES:

Most operating parameters, counters, and pointers are stored in the internal CPU registers. There are three simple data structures:

1. STACK. Used for temporary storage during computations and for holding data to be output with the OUT instruction.

2. PARAMETER STRINGS. There may be one to eight six-byte strings. Each string consists of three 16 bit (two byte) values stored high byte first: number of samples, interval, and delay. A value of zere samples indicates the end of the string area.

3. DATA VALUES. Two-byte data values are stored high byte first, beginning directly above the last parameter string and continuing up through the highest location in the available memory.

TIMING:

All TREC timing routines are based on a 38.400 kHz CPU clock. Assuming this clock frequency, DLAY and AQUI will execute their respective delay and sample intervals with an accuracy of .005%. At the beginning of each new sample set there will be an uncompensated delay of 16 milliseconds while the new sample parameters are accessed.

PROGRAMMER:

Ed Mellinger

DATE:

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25 April 1982

Appendix 1

AVRG:

The program AVRG is a modified version of TREC designed for use when noise is a problem in the A/D converter input. AVRG operates in a manner similar to TREC during all modes of operation, with the exception of:

INITIALIZATION. The INTERVAL value should be specified in <u>seconds</u> rather than in tenths of seconds as with TREC. A value of 1 (1 sec) is acceptable to AVRG, but not to TREC.

The maximum total number of samples in all sample sets is 1325 decimal (52D hex).

AQUISITION. The symbol "S" is output once per second while in aquisition mode. After one specified sample interval (1 to 65536 sec) has elapsed, a set of eight samples is aquired over a period of one second and averaged to produce a single value. This value is stored in data memory in the usual fashion.

The following minor irregularities have been noted in the operation of AVRG:

1. The sample interval is actually 1.0023 times the specified sample interval, in seconds.

2. The first sample following a delay period will be erroneous due to the A/D converter turn-on time constant.

SOURCE CODE FOR TEMPERATURE RECORDER TREC

AVOCET SYSTEMS 1882 CROSS-ASSEMBLER - VERSION 1.48

SOURCE FILE NAME: TREC.ASH

:DSDP CORING TOOL TEMPERATURE RECORDER OPERATING SOFTWARE

;ED MELLINGER 11 APRIL 1982

:UPDATES: 21 APRIL 1982

6696

;"TREC" IS DMA LOADED INTO LOWEST 600 BYTES OF MEMORY. CPU ;RESET CAUSES JUMP TO "INIT" WHICH INITIALIZES CPU REGISTERS AND ;PROMPTS FOR SAMPLE-SET PARAMETERS. UP TO 8 SAMPLE SETS MAY BE ;SPECIFIED, EACH WITH INDEPENDENT TURN ON DELAY (0-65536 SEC), ;SAMPLE INTERVAL (0-6553.6 SEC) AND NUMBER OF SAMPLES (1-1400). ;MAX TOTAL NUMBER OF SAMPLES IS 1400. ENTERING A VALUE OF ZERO ;SAMPLES TERMINATES "INIT" AND CAUSES "VERIFY" TO BE EXECUTED. ;"VERIFY" PRINTS THE SAMPLE-SET PARAMETERS ENTERED DURING "INIT" ;AND OUTPUTS AN 8 BIT LINEAR CHECKSUM OVER THE OPERATING PROGRAM ;AREA. IT THEN PROMPTS FOR USER ACTION; ENTERING "G" CAUSES ;PROGRAM "MAIN" EXECUTION TO BEGIN, ANY OTHER CHARACTER CAUSES A ;RESTART AT "INIT".

9999

;"MAIN" READS THE SAMPLE-SET PARAMETERS STORED IN MEMORY BY ;"INIT" AND INITIALIZES THE SAMPLE COUNT, INTERVAL TIMER, AND ;DELAY WORKING REGISTERS. "MAIN" THEN JUMPS TO "DLAY" WHICH ; DELAYS FOR THE SPECIFIED TIME; DURING DELAY THE TILT SWITCH ;OPTION IS ACTIVE AS SPECIFIED BELON. AFTER DELAY "DLAY" JUMPS TO ;"AQUI" WHICH POWERS UP THE A/D CONVERTER AND AQUIRES THE SPECIFIED NUMBER OF SAMPLES. "AQUI" CALLS SUBROUTINE "TSEC" TO :IMPLEMENT THE SPECIFIED SAMPLE INTERVAL AND "GETDAT" TO STORE ; AQUIRED DATA INTO THE DATA MEMORY AREA. WHEN THE SPECIFIED "NUMBER OF SAMPLES HAVE BEEN AQUIRED "AQUI" RETURNS TO "MAIN" ; WHERE THE NEXT SET OF SAMPLE-SET PARAMETERS IS READ. IF NUMBER OF SAMPLES IS ZERO THEN AQUISITION TERMINATES AND THE INSTRUMENT ; IS POWERED DOWN WITH AN "INP 7" COMMAND.

;AFTER INSRUMENT RECOVERY THE CPU IS RESET AND BEGINS ;EXECUTION FROM LOCATION 0000. NO-OP INSTRUCTIONS PLACED AT 0000-;0002 BY "MAIN" CAUSE EXECUTION TO TRANSFER TO "PLAYBK" WHICH ;OUTPUTS THE STORED DATA AS A SERIES OF RECORDS OF FOUR HEX ;CHARACTERS EACH. AFTER THE LAST DATA RECORD IS OUTPUT "PLAYBK" ;OUTPUTS THE MESSAGE "STOP" FOLLOWED BY ETX,LF,CR, AND THEN POWERS ;DOWN THE RECORDER. DATA MAY BE OUTPUT AGAIN BY RESETTING THE ;CPU.

0000

0000

;THE MAINLINE ROUTINES "INIT", "VERIFY", "MAIN", "DLAY", ;"AQUI", AND "PLAYBK" ALL EXECUTE WITH PC=0. THEY CALL THE ;UTILITY ROUTINES "INDAT", "OUTDAT", "OUTCHR", "TSEC", "GETDAT", ;AND "OUTNUM" TO IMPLEMENT VARIOUS I/O AND TIMING FUNCTIONS. EACH ;UTILITY ROUTINE HAS A SEPARATE PC AND IS CALLED BY THE "SEP PC" ;TECHNIQUE. TREC

8888

;ROUTINES "INIT", "VERIFY", "INDAT", AND "OUTCHR" ARE LOADED ;INTO THE DATA AREA OF MEMORY. THEY ARE EXECUTED ONCE DURING ;INITIALIZATION AND ARE OVERLAID BY DATA DURING THE ACTUAL ;AQUISITION PROCESS. ROUTINES "PLAYBK", "MAIN", "DLAY", "AQUI", ;"GETDAT", "TSEC", AND "OUTDAT" ARE LOADED INTO LOW MEMORY AND ARE ;ALWAYS PRESENT.

AVOCET SYSTEMS 1802 CROSS-ASSEMBLER - VERSION 1.40

trec Equates

0860

88C4 8848

8847

CAPGEE EQU

'G'

, * * * * * * * * *	******	XXXXX	REGISTER USAGE XXXXXXXXXXXXXXXXXXXXXXXXXXXXX
; Program	COUNTERS	3	
;MAINPC E	EQU F	28	MAINLINE ROUTINES PC
TSECPC E	EQU F	81	"TSEC" PC
INPC E	IUU	23	"INDAT" PC, SHARED WITH "OUTNUM"
OUTDPC B	EQU F	₹4	"OUTDAT" PC
GETPC E	equi i	75	*GETDAT* PC
OUTCPC E	:QU	76	"OUTCHR" PC
OUTNPC I	egu i	7 3	"OUTNUM" PC, SHARED WITH "INDAT"
DATPTR (EQU I	R7	DATA MEMORY POINTER, ALSO "LINSUM" POINTER
PARPTR I	EQU (88	SAMPLE-SET PARAMETER STRING POINTER
STKPTR I	equ i	82	STACK POINTER
ENDFLG I	EQU I	RD	END-OF-DATA-AREA FLAG; SHARED WITH "SMPCTR"
BUFFERS	and wor	KING ST	ORAGE
I ODAT	EQU	R9	DATA FROM/TO I/O ROUTINES
DLYSTR	EQU	Ra	DELAY STORAGE FOR "DLAY"
INTSTR	EQU	R9	CURRENT SAMPLE INTERVAL FOR "AQUI", SHARED WITH "OUTCHR"
;COUNTER	S		
TSCTR	EQU	RB	"TSEC" INTERVAL COUNTER
DLYCTR	EQU	RC	"DLAY" INTERVAL COUNTER
SHPCTR	EQU	RD	"AQUI" SAMPLE COUNTER; SHARED WITH "ENDFLG"
TLTCTR	E8U	RE	"DLAY" TILT SWITCH COUNTER
;SCRATCH	Pad		
;scra	EQU	RF	SCRATCHPAD REGISTER
*****	******	XXX E	QUATES AND CONSTANTS XXXXXXXXXXXXXXXXXXXXXXXXXX
1/0 POR	(T NAMES		
:HBYTE	EQU	INP1	HIGH BYTE OF A/D CONVERTER
LBYTE	EQU	INP2	LON BYTE OF A/D
ADSTRT	EQU	INP3	a/d start signal
UARTIN	EQU	INP4	UART REVR BUFFER REGISTER
UARTOT	EQU	OUT5	UART XMITTER BUFFER REGISTER
UNUSED	EQU	INP6	UNUSED I/O SLOT
POWRDN	EQU	INP7	POWER DOWN
; CONSTAL	VTS		
NOINST	EQU	0C4H	;OP-CODE OF "NOP" INSTRUCTION
DLAYMK	EQU	046H	:CHAR "&" FOR "DLAY" TIME MARK

USER GO SYMBOL

trec Equates

006A	LF	EQU	8AH	;LINEFEED CHARACTER
000D	CR	EQU	ØDH	CARRIAGE RETURN
8828	SP	EQU	828H	SPACE
0003	ETX	EQU	63H	END OF TEXT CHARACTER

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TREC PROGRAM

PROGRAM "TREC" ţ

0000 C00131	LBR	INIT	;ON INITIAL CPU RESET,
0003			JUMP TO "INIT".
0003			;ON LATER RESET,
9983			;JUMP TO *GLITCH".
0003			;ON LAST RESET,
8863			;FALL THRU TO "PLYBK"

;ENTERED ON CPU RESET AFTER FIRST THREE LOCATIONS ARE FILLED WITH NOPS BY "MAIN". USES "PARPTR" AS DATA POINTER, LOADS EACH TWO BYTE SAMPLE INTO 'IODAT' AND CALLS 'OUTDAT' TO XMIT DATA VIA WART. EACH 4 DIGIT ASCII HEX VALUE IS FOLLOWED BY LF, CR, AND A ;0.2 SEC PAUSE. RECORDER IS POWERED DOWN AFTER END OF DATA.

0003 E2 SEX R2 FIX STACKPTR AFTER RESET

AT TERMINATION OF "MAIN", "PARPTR" POINTS TO START OF DATA AREA. "TRANSFER IT TO "DATAPTR" FOR USE BY "PLAYBK".

0984	98	GHI	R8
0005	B7	PHI	R7
6663	88	GLO	R8
0007	A7	PLO	R7

; COMPARE "DATPTR" TO "ENDFLG" TO CHECK FOR END OF DATA

8999	97	PLAYBK:	GHI	R7	;get data ptr
0007	52		STR	R2	PUT ON STACK, NO PUSH
000A	90		GHI	RD	GET FLAG
886B	F7		SM		SUBTRACT FLG - PTR
000C	3A14		BNZ	Moudat	;IF .NE. DO DATA OUTPUT
000E	87		GLO	R7	;ELSE CHECK
966F	52		STR	R2	LOW BYTES ALSO
0010	8D		GLO	RD	
6011	F7		SM		; IF LB'S MATCH,
0012	3223		BZ	PBEND	GO END PLAYBACK
0014					ELSE NEXT DATA PT.

;MOVE DATA TO "IODAT", SEND WITH "OUTDAT", SEND LF, CR WITH "OUTCHR"

0014 47	MOVDAT: LDA	R7	;MOVE 2 BYTES
0015 B9	PHI	R9	;(1 sample) into
0016 47	LDA	R7	;I/O BUFFER
8817 A9	PLO	R9	AND SEND WITH
0018 D4	SEP	R4	;CALL "OUTDAT"

; "OUTNUM" -- DEDICATED SUBROUTINE FOR "OUTDAT". CONVERTS 4 BIT BINARY ;VALUE IN D REG TO ASCII HEX (8-9, A-F) AND XMITS VIA WART.

862E	D9	OTEXIT:	SEP	RØ	RETURN TO CALLER
902F	99	OUTDAT:	GHI	R9	HIGH BYTE TO XMIT
8839	Fó		SHR		
8831	F6		SHR		
0032	Fó		SHR		
0033	F6		SHR		ISOLATE HIGH NIBBLE
0034	D3		SEP	R3	SEND IT
0035	99		GHI	R9	HIGH BYTE
0036	FAOF		ANI	8FH	LOW NIBBLE
00 38	03		SEP	R3	;SEND IT
8839	89		GL0	R9	;LON BYTE
883A	F6		SHR		
663B	F6		SHR		
903C	F6		SHR		
863D	F6		SHR		HIGH NIBBLE
003E	D3		SEP	R3	SEND IT
683F	89		GLO	R9	LOW BYTE
8848	FAOF		ANI	ØFH	LOW NIBBLE
0942	D3		SEP	R3	SEND IT
6043	302E		BR	OTEXIT	RESET "OUTDPC" BEFORE EXIT

; "OUTDAT" TRANSMITS CONTENTS OF REGISTER 'IODAT' AS 4 ASCII HEX DIGITS. ; "OUTDAT" ENTERED VIA "SEP OUTDPC" AND EXITS VIA "SEP MAINPC".

;CALL "OUTCHR" TO

002E XXXXXXXXXXXXXXXXXXXXXXXXX	outdat/outnum	*****
--------------------------------	---------------	-------

RØ

00 1A	8A9D89		DB	LF,CR,0	;SEND LF,CR
88 1D 90 1F 8929	F802 A B D1		ldi Plo Sep	2 R9 R1	;DELAY 0.2 SEC ;BY CALL TO ;"TSEC"
0021	3098		BR	Playbk	;KEEP SENDING DATA
0023 0024 0028	D6 53544F59 830A8D 88	PBEND:	sep DB DB	R6 'STOP' ETX,LF,CR,0	; SEND MSG. WHEN THRU
002C 002D	6F 88		INP7 IDL		;Then power down ;AND IDLE FOR SAFETY

R6

TREC PROGRAM

8819 D6

862E D0

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SEP

OTEXIT: SEP

TREC

program

6045	D4	ONEXIT:	SEP	R4	RETURN TO "OUTDAT"
9946 9946	FF9A	outnum:	SMI	ØAH	TEST FOR <= 9
0049 0049	FC07		ADI	7	; IF A-F, ADD EXTRA
004B	FC3A		ADI	03AH	;ADD NET 39 OR
884D 994D					;NET 36 TO GET :ASCLI CHARACTER.
004D	52		STR	R2	PUT ON STACK, NO PUSH
004E	344E	ONLOOP:	81	ONLOOP	WAIT FOR UART TRE
8859	65		0075		SEND TO WART
0051	22		DEC	R2	;UNDO AUTOINCREMENT
0052	3045		BR	ONEXIT	;RESET "OUTNPC"

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;"MAIN" READS SAMPLE-SET PARAMETERS FROM MEMORY (POINTED TO BY ;(PARPTR") AND STORES THEM IN "SMPCTR", "INTSTR", AND "DLYSTR", ;IF SPECIFIED NUMBER OF SAMPLES IS ZERO, AQUISITION STOPS AND THE ;RECORDER IS POWERED DOWN. IF # SAMPLES IS NONZERO "MAIN" JUMPS ;TO "DLAY" FOR DELAY BEFORE AQUISITION BEGINS. AFTER DELAY, ;"DLAY" JUMPS TO "AQUI" WHERE THE SPECIFIED NUMBER OF SAMPLES ARE ;AQUIRED, FOLLOWED BY A RETURN TO "MAIN" WHERE THE NEXT SAMPLE-;SET'S PARAMETERS ARE READ.

0054 ;MAIN, DLAY, AND AQUI EXECUTE WITH PC=MAINPC=R0. IN ;GENERAL, THETIMING OF THESE ROUTINES IS CRITICAL TO OVERALL ;TIMING ACCURACY. ANY PROGRAM CHANGES SHOULD THEREFORE BE MADE ;WITH CARE.

0654	48	MAIN	lda	R8	;get sample hb
0055	BÐ		PHI	RD	;STORE
0056	52		STR	R2	;ALSO SAVE TEMP.
0057	48		lda	R8	;GET # SAMPLES LB
8858	AD		PLO	RD	;store
0059	F1		OR		;merge lb and HB
085A	3A6C		BNZ	NOIDLE	;KEEP GOING IF<>0
995C					;ELSE POWER DOWN

;BEFORE POWERING DOWN PUT "NO-OPS" AT 0000-0002, TO ALLOW ;"PLAYBK" TO BE EXECUTED AFTER NEXT CPU RESET. (NOTE D=0 HERE)

062C	BF	PHI	RF	;FIRST NEED
005D	AF	PLO	RF	;POINTER TO 8000
005E	F8C4	LDI	NOINST	;GET "NOP" OPCODE
0860	5F	STR	RF	;PUT AT 0000
0061	iF .	INC	RF	
99 62	5F	STR	RF	;AND AT 8001
8963	1F	INC	RF	

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TREC

Program

8864	5F	STR	₽F	;AND AT 8882
		;al so sa ve last	Data	Location as flag for *Playbk*
0065	97	GHI	R7	
8866	BD	PHI	RD	
8867	87	GLO	R7	
8068	AD	PLO	RD	
8069	7A	REQ		:Turn off a/d
996A	6F	INP7		POWER DOWN
696B	90	IDL.		; IDLE TO BE SURE
		;IF # SAMPLES C	> 9 Gł	et interval and delay parameters also
906C	48	NOIDLE: LDA	R8	;GET AND STORE
006D	B9	PHI	R9	SAMPLE INTERVAL
006E	48	LDA	R8	HB THEN LB.
906F	A9	PLO	R9	,
0870	48	LDA	R 8	SET AND STORE
	 .			,

 0071
 BA
 PHI
 RA
 ;DELAY TIME

 9072
 48
 LDA
 R8
 ;HB THEN LB.

 9073
 AA
 PLO
 RA

;FALL THRU INTO "DLAY" WHEN ALL PARAMETERS ARE LOADED. "DLAY" ;MOVES "DLYSTR" INTO "DLYCTR", DECREMENTS "DLYCTR" ONCE PER ;SECOND UNTIL "DLYCTR" IS ZERO, THEN JUMPS TO "AQUI". TILT ;SWITCH IS SAMPLED EVERY SECOND DURING DELAY. IF SWITCH OPENS ;FOR 60 CONSECUTIVE SAMPLES (1 MINUTE) THEN "RESTRT" IS EXECUTED. ;"RESTRT" RESETS DELAY COUNTER TO ITS INITIAL VALUE ("DLYSTR" ;INTO "DLYCTR"), THEN SAMPLES TILT SWITCH EVERY 4 SECONDS. IF ;SWITCH CLOSES FOR 75 CONSECUTIVE SAMPLES (5 MINUTES) THEN "DLAY" ;IS EXECUTED AND DELAY STARTS OVER.

;TIMING IN THESE ROUTINES IS CRITICAL; MAKE CHANGES CAREFULLY.

8874	9A	DLAY:	GHI	RA	GET DELAY TIME
9675	BC		PHI	RC	FROM STORAGE
8876	8A		GLO	RA	; INTO
9677	AC		PLO	RC	WORKING COUNTER.
0078	F83C		LDI	03CH	;68 SEC TIMEOUT
997A	AE		PLO	RE	TO DETECT TILT.

;("GLITCH" IS THE RE-ENTRY POINT IN CASE OF A SPURIOUS CPU RESET ;DURING "DLAY" OR "AQUI"; IT IS ENTERED AFTER CPU RESET VIA A ;LONG BRANCH PLACED AT 0000-0002 BY "INIT". RECOVERY FROM ;SPURIOUS RESETS IS POSSIBLE DURING "DLAY" AND "AQUI", WITH ;POSSIBLE LOSS OF ONE DELAY OF AQUISITION CYCLE. GLITCHES DURING ;MAIN" ARE FATAL SINCE "PARPTR" WILL BE INCORRECTLY POSITIONED

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trec Program

;UPON RE-ENTRY.)

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907B	E2	GLITCH:	SEX	R2	;FIX R(X) AFTER RESET
		;TEST FO	or delay	FINISHED; TEST F	FOR TILT
907C	90	DYLCOP:	GHI	RC	;TEST DELAY
807D	52		STR	R2	BY MERGING
887E	80		GLO	RC	DLYCTR HB+L8.
687F	FI		OR		IF BOTH ZERO
0080	32B4		BZ	AQUI	;60 DO AQUI.
0082	8E		GLO	RE	ELSE TEST FOR
8083	32A2		BZ	RESTRT	TILY, RESTART IF YES
		;OUTPUT	*0* For	TIMING, DELAY O	WE SECOND, CHECK TILT SWITCH
0085					;IF ALL OK,
8685	6E		INP6		DO TIME MARKS
6686	F840		LDI	DLAYMK	XMIT MARK VIA
6088	52		STR	R2	UART EVERY SEC.
8889	45		01075		,
688A	22		DEC	R2	;UNDO AUTOINCR.
008B	20		DEC	RC	;DEC DELAY COUNTER REG
2000	C000		זמו	0	COARSE DELAY
000U	F007 AD			/ 00	A O CEP UAT
0005	HD D4		CED	07	JULY DEL VHI
669r	וע		3EF	K1	; (366 GHLL.
8898	F849		LDI	849H	;FINE DELAY (49H=73D)
8892	FF61	DELOOP:	SMI	1	TO FILL OUT DYLOOP
0094	7A		REQ		(ALSO OFF A/D)
0095	3492		BNZ	DELOOP	TO EXACT 1.0 SEC
AR97	3E9E		RN4	DTILT	CHECK TILT SM.
8899	F83C		IDT	93CH	IF NO TILT.
0077	ΔF		Р. П	RF	PRESET THE COUNTER
ANAC	3970		8R	DYL DOP	KEEP DELAYING.
0070	0010				,
009E	2E	DTILT:	DEC	RE	;IF TILT, DEC COUNTER.
009F	7A		REQ		;(2 CYCLE NO-OP)
9880	307C		BR	DYLCOP.	;KEEP DELAYING.
		;"RESTR ;TILT S	T" RESE WITCH TO	ts the delay cou Open and stay o	nter ("Dlyctr") and waits for Pen 5 minutes.
0047	EOND	DECTOT .	וחו	0 A DU	1404-750 EAD
OUHZ	7040 AC	REDIRI		9400 00	JADD-130 FUR
0644	HE.		rlu	RE	jo nan kesinki delat.

00A5 8E	RSLOOP: GLO	RE	TEST FOR UNTILT

.

THE
TREC

Program

88A6 3274	BZ	DLAY	;re-start delay if yes
86A8 F828	LDI	928H	;4.9 SECOND
00AA AB	PLO	RB	COARSE DELAY ONLY
88A8 D1	SEP	R1	VIA "TSEC" CALL.
88AC 3781	B4	NOTILT	;TEST TILT SW
88AE F84C	LDI	04CH	; IF OPEN, RESET
00B0 AE	PLO	RE	;TILT COUNTER (+1)
90B1 2E	NOTILT: DEC	RE	; IF CLOSED, DEC
0082 30A5	BR	RSLOOP	COUNTER AND LOOP.

; "AQUI" IS ENTERED FROM "DLAY". "AQUI" CHECKS SAMPLE COUNTER ;REG. "SMPCTR" FOR 2ERO, RETURNS TO MAIN IF YES. IF NO, "AQUI" ;POWERS UP A/D, INITIALIZES "TSEC" COUNTER WITH SAMPLE INTERVAL ;AND CALLS "TSEC". AFTER "TSEC" TIMES THE SPECIFIED INTERVAL ; "AQUI" STARTS A/D CONVERSION AND CALLS "GETDAT" TO TRANSFER ;CONVERTER DATA TO MEMORY IN DESIFED FORMAT (PACKED OR UNPACKED). ;AFTER TRANSFER, "AQUI" CHECKS DATA POINTER "DATPTR" AND PREVENTS ;IT FROM INCREMENTING BEYOND THE END OF DATA MEMORY (LOCATION ;0BFF).

;"AQUI" IS A MAINLINE ROUTINE (PC=MAINPC=R0). ITS TIMING IS ;CRITICAL AND MODIFICATIONS SHOULD BE MADE WITH CARE.

FIRST CHECK FOR DONE

00B4 9D ·	AQUI :	GHI	RD	;GET SAMPLE COUNTER HB
8685 52		STR	R2	PUT ON STACK, NO PUSH
00B3 8D		GLO	RD	GET SAMPLE CTR LB
8087 F1		OR		MERGE HB+LB
8088 3254		8Z	MAIN	IF BOTH ZERO, GO DO
00BA				NEXT SAMPLE SET.

; IF NOT DONE, SEND MARK AND WAIT I SAMPLE INTERVAL

80BA	7B		SEC	•	;PWR UP A/D
00BB	бE		INP6		PULSE I/O 6 LINE
88BC	2D		DEC	RD	DEC SAMPLE COUNTER
96BD	99		GHI	R9	MOVE INTERVAL
00BE	8B		PHI	RB	INTO "TSEC" WRKING
00BF	89		GLO	R9	COUNTER
8909	AB		PL0	RB	
00C1	2B		DEC	RB	LESS 0.1 SEC FOR
88C2					LATER FINE DELAY.
88C2	Di		SEP	R1	DELAY SPEC'D TIME
89C3	бB		INP3		START A/D CONVERSION
98C4	F86A		LDI	96AH	FINE DELAY
00C6	FF81	AQLOOP:	SHI	1	WAIT FOR A/D

TREC

PRC

PROGRA	м					
86C8	3AC6		BNZ	AQLOOP		
		;GET DAT	a from 1	a/d and check f()r data	Memory Full
00CA 00CB	D5		SEP	R5	;call ;move	"getdat" to data to mem
00CB 00CC 00CE	97 FF0C 3BB4		ghi Smi Bm	R7 Och Aqui	; Check ; Data ; Loop	(FOR MEM FULL IF NO
00D0 00D1 00D2	27 27 3054	; (THIS ; Sample ; For Pla	dec dec 9r Allows Parame Yback.)	R7 R7 MAIN CORRECT TERMINA TERS MUST BE REJ	; IF YE ; DEC F ; AND F TION OF AD TO CO	ES, POINTER FORCE NEXT AQUISITION IN "MAIN". ALL DRRECTLY POSITION "PARPTR"
		; "getdat ; sequent ; samples	r" N TIALLY 3 IF 12-	onpacking vers In Memory. () BIT Samples are	ION. (PACKING USED),	gets a/d data and stores Version Allows 33% more
80D4 00D5 00D6 00D7 00D8 00D8 00D9 60D4	D8 69 57 17 6A 57 17	GDEXIT: GETDAT:	SEP INP1 STR INC INP2 STR INC	R0 R7 R7 R7 R7	;RETUI ;GET ;PUT ;INC ;GET ;PUT ;INC	RN TG "Aqui" A/d HB In Data Mem Pointer A/d LB In Data Mem Pointer
8809	3804		BR	GDEXIT	;REST	ORE "GETPC"
		;*****	*****	*******	NUTCHR	*****
88DD	1	;POINTE ;Return	;enteri d to i s via "	ed VIA "Sep ol By "Mainpc" Unti Sep Mainpc". Al	ΠCPC". IL A 00 .TERS MC	OUTPUTS CHARACTER STRING BYTE IS ENCOUNTERED, THEN STKPTR).
80DD 00DE 80DF 80E 1) D 9 40 32DD 52	OCEXIT: OUTCHR:	sep LDA BZ Str	RØ RØ OCEXIT R2	;RETU ;GET ;IF 0 ;ELSE	RN TO CALLER CHARACTER 10, EXIT : PUT ON STACK
08E2 08E4 00E5 00E5	2 34E2 1 65 5 22 5 30DE	OCLOOP:	B1 OUT5 DEC BR	ucloup R2 Outchr	;WAIT ;XMIT ;UNDC ;GET	HUR UART WHEN READY) AUTOINCR. NEXT CHAR
		;*****	******	****	TSEC	************

;"TSEC" IS A 0.1 SEC TIMING ROUTINE. TIME FROM BEGINNING OF ;EXECUTION OF FIRST INSTRUCTION (DEC TSCTR) TO END OF EXECUTION

TREC

PROGRAM

00E8		;of las ;assumi) ;call W	T INSTRU ;(0.1) NG A 38. ITH "SEP	NCTION (S X (VALUE 400 KHZ P TSECPC"	ep ma In CPU c Re	inpc) TSCTR" Lock. Turn T	IS > SECONDS, O MAINPC (: (R9) .		
80E8	D8	TSEXIT:	SEP	RØ		;R	eturn to (ALLE	R	
90C7 90F4	20 F974	1955	IDEC	RB 074U		μ 	EL LUUNIEN ET UD I OCA	(00	
0041	IVII			0710		30	EI OF LUGF	nt tu	UF	
00EC	FF01	TSLOOP:	SMI	t						
90EE	3aec		9NZ	TSLOOP						
AAFA	98		GHT	99		•0	HECK COLMO	.ED		
AAFI	52		STR	R0 R2		ιD	חבטה טטוח וזד עס האו מ	בת דהרע		
AAE2	88		61 A	ממ		۱۴ ۱۵	on no un c ETTID	HON		
AAF3	F1 '		0CU NR	ΝQ		ju •M	ENCE VRAIG	2	-	
68F4	32E8	,	RZ	TSEXIT		دىر F:	XIT IF RAT	, ห.ด.		
•						, <u> </u>	(1) II QQI	13 Y		
00F6	30E9	ENDCHK:	BR	TSEC		şΕ	LSE KEEP L	.00P1	NG	
		;("ENDCI	4K" IS /"	LABEL	FOR	LAST	LOCATION	TO	CHECKSUM	DURING

8108	ORG	\$+8	;SMALL, STACK AREA
0101	STKTOP: ORG	\$+1	• ,
0131	Param: Org	\$+48	SPACE FOR
6131			8 SAMPLE-SET PARAMETER STRINGS

;EACH SAMPLE-SET PARAMETER STRING IS 6 BYTES LONG; ;TWO BYTES EACH (HB FIRST) FOR: # SAMPLES, SAMPLE INTERVAL, AND ;TURN-ON DELAY. DURING AQUISITION, SATA IS STORED IN MEMORY ;BEGINNING DIRECTLY AFTER (ABOVE) THE LAST PARAMETER. THE LAST ;PARAMETER IS ALWAYS THE "ZERO SAMPLES" FLAG WHICH TERMINATES ;INPUT AND AQUISITION.

;THE FOLLOWING ROUTINES ARE LOADED INTO THE DATA AREA AND ARE ;EXECUTED ONCE DURING INITIALIZATION. THEY ARE OVERLAID ;(DESTROYED) BY DATA DURING THE ACTUAL AQUISITION PROCESS.

0131

;ENTERED VIA LONG BRANCH INSTRUCTION AT 0000-0002, AFTER ;CPU RESET. INITIALIZES REGISTERS TSECPC, INPC, OUTPC, OUTCPC, ;GETPC, STKPTR, PARPTR; SETS X=STKPTR; POWERS DOWN A/D. PROMPTS TREC

PROGRAM

;USER FOR INPUT OF SAMPLE-SET PARAMETERS: ; #SAMPLES? # OF SAMPLES IN THIS SET; ZERO STOPS FURHTER INPUT. ; INTERVAL? SAMPLE INTERVAL FOR THIS SET, IN TENTHS OF SECONDS. ; DELAY? TURN ON DELAY BEFORE AQUIRING THIS SET, IN SECONDS.

;THE RESPONSE TO EACH PROMPT SHOULD BE A ONE TO FOUR CHARACTER ;HEX INTEGER FOLLOWED BY A CARRAIGE RETURN. THE INPUT PROTOCOL ;IS UT4 FORMAT, I.E. ONLY THE LAST 4 DIGITS ARE SIGNIFICANT; ;THUS ERRORS MAY BE CORRECTED BY TYPING UNTIL THE NUMBER IS ;CORRECT AND THEN TYPING "RETURN". ILLEGAL OR UNRECOGNIZED ;CHARACTERS ARE NOT ECHOED BACK TO THE USER INPUT DEVICE.

- 0131 ;"INIT" CONTINUES TO PROMPT FOR SAMPLE-SET PARAMETERS ;UNTIL A VALUE 0 IS READ FOR # SAMPLES. UP TO 8 SETS OF ;PARAMETERS MAY BE ENTERED; MORE WILL OVERLAY PROGRAM AREAS. ;"INIT" EXITS WITH A JUMP TO "VERIFY", KEEPING PC=MAINPC=R0.
- 0131 7A / INIT: REO ;TURN OFF A/D

;INITIALIZE WORKING REGISTERS AS REO'D

8132	F899	LDI	A.1(TSEC)	;INITIALIZE
0134	B1	PHI	R1	; TSEC PC.
0135	F8E9	LDI	A.8(TSEC)	
0137	Al	PLO	R1	
A 120	5992	101		
0100 0100	02	PHT	P3	. TNDATE PC.
0100	00 CORA			y 110011 1 0 1
0100	FOUR AD		010011110011	
0130	нэ	FLU	ĸə	
013E	F800	LDI	A.1(OUTDAT)	
0140	B4	PHI	R4	;"OUTDAT" PC.
0141	F82F	LDI	A.8(OUTDAT)	
6143	A4	PLO	R4	
Q 144	F898	IDT	A. 1(0)(TCHR)	
01/14	P.C.	PHI	RA	. "OUTCHR" PC.
0140	CODE			3 0010100 101
0140	F 00/E		D2	
0147	но	FLU		
814A	F800	LDI	A.1(GETDAT)	
014C	B5	PHI	R5	;"GETDAT" PC.
614D	F8D5	LDI	A.9(GETDAT)	•
014F	A5	PLÛ	R5	
0150	E00 (Int		
0100	1001	011	D11119101917	TOTACY DOINTED
0102	D4 .	ពារ ក្រារ	R£ A 0/071/700\	JOINUN FUINTER
0103	1000		H.0(31K)UP/	
6122	AZ	FLU	RZ	

TREC PROGRAM

8177 58

0178 52

0179 18

917A 89

0178 58

017C 18

817D F1

0180

017E 32AD

0156 0158 0159 0159 0158	F801 88 F801 A8		LDI PHI LDI PLO	A.1(PARAM) R8 A .9(PARAM) R8	;Parameter ;String pointer
815C	ε2		SEX	R2	;X=STKPTR ALWAYS
		; prompt	USER FO	r sample -set par <i>i</i>	METERS
015D 015E	06 8a0 0 80	PROMPT :	sep Db	R6 LF,CR,0	;CALL "OUTCHR" TO ;SEND LF,CR.
8161 8163 8164 8166 8167	F890 BB F802 AB D1		LDI Phi LDI Plo Sep	9 RB 2 RB R1	;SET UP 0.2 SEC ;DELAY AFTER CR ;CALL "TSEC"
		;prompt	and get	# of samples	
8168 8169 8171	D6 2353414D 3F208308		sep DB DB	R6 '#SAMPLES' '?',SP,ETX,8	;call "Outchr"
0175 0176	D3 99		SEP GHI	R3 R9	;CALL "INDAT" ;GET RETURN HB

; PROMPT AND GET SAMPLE INTERVAL

STR

STR

INC

GLO

STR

INC

ÛR

BZ

R8

R2

R8

R9

R8

R8

BURN

0180 D6 0181 202020 0184 49455445 018C 3F200300	SEP DB DB DB	R6 SP,SP,SP ′INTERVAL′ ′?′,SP,ETX,0	;send prompt
0190 D3	SEP	R3	GET INPUT
8191 99	GHI	R9	·
8192 58	STR	R8	
8193 18	INC	R8	
6194 89	GLO	R9	
8195 58	STR	R8	

.

STORE IN MEMORY

;ALSO ON STACK

;GET RETURN LB

; INC MEM PTR

STORE IN MEMORY

;MERGE HB AND LB

; IF 0, NO MORE INPUT.

;ELSE GET INTVL, DELAY.

;INC MEM PTR

trec Program

LUQUULI

0196 18 INC

; PROMPT AND GET TURN ON DELAY TIME (DELAY BEFORE SAMPLING STARTS)

8197 9198 9198 9198	Dő 292929 44454C41 299399	SEP DB DB DB	R6 SP,SP,SP 'DELAY?' SP FTY A	;send prompt
0141	200300	V Q	JEINJO	
0 1A4	D3	SEP	R3	GET INPUT
01A5	99	GHI	R9	STORE IN
0 1A6	58	STR	R8	MEMORY
0 1A7	18	INC	R8	PARAM STRING AREA
0 1A8	89	GLO	R9	
01A9	58	STR	R8	
0 1AA	18	INC	R8	
8 1AB 8 1AD	305D	BR	PROMPT	;GET NEXT ;PARAMETER SET

R8

;CHANGE "LONG BRANCH" INSTRUCTION AT 0000-0002 TO RE-ENTER "DLAY" ;IN CASE OF SPURIOUS RESET.

0 1AD	F800	BURN:	LDI	8	;SET UP POINTER
8 IAF	BF		PHI	RF	TO LOCATION
8180	AF		PLO	RF	
0181	1F		INC	RF	;of Jump Address
81B2	F869		LDI	A.1(GLITCH)	GET NEW ADDRESS
81B4	5F		STR	RF	STORE IN 8001
0 i B5	1F		INC	RF	
9186	F87B		LDI	A.8(GLITCH)	
0188	5F		STR	RF	;AND IN 8002

FALL THRU TO VERIFY WHEN THRU

0 i B9

;OUTPUTS # SAMPLES, INTERVAL, AND DELAY DATA FOR USER ;INSPECTION. COMPUTES LINEAR CHECKSUM (LINEAR SUM OF MEMORY ;BYTES) OVER PROGRAM AREA FROM 0000 TO LABEL "ENDCHK" AT END OF ;"TSEC", AND OUTPUTS RESULT. PROMPTS WITH "ACTION...?" FOR USER ;RESPONSE. ENTER UPPERCASE "G" TO BEGIN EXECUTION OF "MAIN", ANY ;OTHER CHARACTER TO START "INIT" OVER AGAIN. "PARPTR" AND ;"DATPTR" ARE RE-INITIALIZED BEFORE THE JUMP TO "MAIN".

#189 ; "VERIFY" EXECUTES WITH PC=MAINPC=R0.

0189 F880	VERIFY: LDI	A.1(OUTNUM)	;FIRST BUSINESS IS
91BB 83	PHI	R3	TO INIT "OUTNPC"
01BC F846	LDI	A.8(OUTNUM)	NOW THAT "INDAT"

TREC

PROGRAM

8 1BE	A3		PLO	R3	;IS FINISHED
0100	C00 1		LDT		DECET DADDTO TO
0100	L001			H+ICCHCHII/	ARDEL PHAFIR (U
DICI	68		PHI	Ka	SIAKI UF
01C2	F891		LDI	A . U (PARAM)	;PARAM STRING
01C4	A8		PLO	R8	;BEFORE VERIFYING
8105	D6		SEP	Ró	:Call Outchr and
81C6	9A9A99		DB	LF,LF,0	ILEAVE THO BLANK LINES
0100	57	ur aan.	ero	D /	J# 000 IC
8167	00	VLUUP:	357		
01LA	RURDAR		DB	LF,CR,0	;001PUT LUUP
91CD	F802		LDI	2	;0.2 SEC DELAY
81CF	AB		PLO	RB	AFTER CR. VIA
A 109	Ð1		SEP	R1	CALL TO "TSEC"
0 100					Junice to toes
		;OUTPUT	#ISAMPLES	; EXIT VLOOP IF	=8.
01D1	48		LDA	R8	:GET #SAMPLES HB
A 1D2	89		PHI	R9	PUT IN OUT BUFFFR
A 1D3	52		STR	82	ALSE IN STACK
0100	10 JL			00	ACT ACANDIEC ID
0104	40			NO DO	JUE! HOHENELED LD
8103	AY		FLU	КУ	PUT IN OUT BUFFER
81D6	F1		or		;merge HB and LB
8 1D7	C2020F		LBZ	LINSUM	;EXIT VLOOP IF 0
81DA	D6		SEP	Ró	ELSE SEND TEXT
A IDR	23534140		DB	14SAMPLES	,
0100	20001110		NØ	/ - / CD 0	
0100	302000			- 35r30	
0 IE6	U4		SEP	K4	THEN SENU DATA
		;output	SAMPLE	Interval	
91F7	49		۱n۵	88	RET INTERUAL HR
0100	00		PHI	po	
01E0	40		1114	N7 D0	JEOL IN DOLLA
8 IEA	48		LUH	NO	JOES INTERVAL LD
91EA	A7		PLU	KA	PUT IN BUFFER
01EB	D6		SEP	R6	;SEND TEXT
0 1EC	202020		DB	SP.SP.SP	
RIFE	49455445		DA	INTERUAL /	
Q1E7	202000		NR NR	/=/ CD ₽	
0104	DN 0000		CED	- jvij0 DA	THEN CENT DATA
@1LH	U 4		JEL	K 7	jinch John Uhth
		;OUTPUT	turn-on	DELAY	
0 1FB	48		lda	R8	GET HB

01 0 10	L L L L L L L L L L L L L L L L L L L	110	
01FC 89	PHI	R9	
01FD 48	LDA	R8	;GET LB
01FE A9	PL0	R9	

•

TREC

Program

0 1FF	06	SEP	R6	;SEND TEXT
0200	202020	DB	SP,SP,SP	
8283	44454C41	DB	'DELAY=',SP,0	
028B	D4	SEP	R4	;THEN SEND DATA
828C	C081C7	LBR	VLOOP	REPEAT FOR NEXT
026F				;parameter set

• •

; COMPUTE CHECKSUM AND OUTPUT RESULT.

020F	F806	LINSUM:	LDI	8	FIRST SET UP
0211	B7		PH1	K/	PUINIER IV
8212	A7		PLO	R7	;program area
0213	89		PHI	R9	;Clear Buffer HB
0214	E7		SEX	R7	;READY FOR ADDING
0215	89	CKLOOP:	GLO	R9	;GET OLD SUM
0216	F4		add		; add memory byte
8217	A9		PLO	R9	SAVE NEW SUM
8218	17		INC	R7	;INC POINTER TO NEXT BYTE.
0219	97		GHI	R7	;TEST IF THRU BY
021A	FF60		SMI	A.1(ENDCHK)	;Comparing PTR
021C	CB0215		LBNF	CKLOOP	AND FLAG LABEL.
021F					;KEEP LOOPING IF NO
821F	87		GL O	R7	;BOTH BYTES
0220	FFF6		SMI	A.8(ENDCHK)	;must match
8222	CB8215		LBNF	CKLOOP	; TO BE DONE
0225	E2		SEX	R2	;re-enable stack
8226	D6		SEP	R6	;WHEN DONE, SEND
8227	20202043		DB	SP,SP,SP,'CHECK'	,
822F	53554D3D		DB	′SUM≓′,SP,8	
0235	04		SEP	R4	;and send data

; OUTPUT PROMPT AND READ GO/NO GO CHARACTER.

	sep D B DB	R6 SP,SP,SP,'ACT' 'ION?',SP,ETX,0	;send prompt
ACLOOP:	B2	ACLOOP	; AWAIT UART REPLY
	INP4		JOEI REPLI
	SHI	Capgee	;IS IT "6" ?
	LBNZ	INIT	;IF NO, "INIT"
			;IF YES, "MAIN"
	ACLOOP:	SEP DB DB ACLOOP: B2 INP4 SMI LBNZ	SEP R6 DB SP,SP,SP,'ACT' DB 'ION?',SP,ETX,0 ACLOOP: B2 ACLOOP INP4 SMI CAPGEE LBNZ INIT

;AFTER LAST SAMPLE-SET PARAMETER HAS BEEN READ "PARPTR" POINTS TO ;START OF DATA AREA. TRANSFER IT OT "DATPTR", THE RE-INITIALIZE ;"PARPTR". THEN JUMP TO "MAIN" TO BEGIN AQUISITION PROCESS.

trec Program

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824C 98	GHI	R8	
024D B7	PHI	R7	
024E 88	GLO	R8	
824F A7	PLO	R7	
0250 F801	LDI	A.1(PARAM)	
0252 88	PHI	R8	
0253 F801	LDI	A.O(PARAM)	
8255 A8	PLO	R8	
8256 C88854	LBR	MAIN	;END OF OPERATOR
0259			; INITIALIZATION

0259

;CALLED VIA "SEP INPC". CLEARS "IODAT", THEN READS RCVD. ;CHARACTERS FROM UART. HEX DIGITS (0-9,A-F) ARE CONVERTED TO ;4 BIT VALUES AND SHIFTED INTO "IODAT" FROM RIGHT TO LEFT, AND ;ALSO ECHOED TO USER. NON-HEX CHARACTERS ARE IGNORED. INPUT ;STRING IS TERMINATED WITH A CR; RETURN IS VIA "SEP MAINPC". ;ASSUMES X=STKPTR.

8259 D8	INEXIT: SEP	RØ	;return to caller
925A F896	INDAT: LDI	0	;ON ENTRY
025C B9	PHI	R9	CLEAR BUFFER
025D A9	PLO	R9	

; INPUT A CHARACTER, CHECK IT, CONVERT IT TO 4 BITS, ECHO IT

025E 0260	355E 6C	ILOOP:	B2 INP4	ILOOP	;WAIT FOR CHARACTER ;THEN GET IT
8261 8263	FF0D 3259		SMI BZ	CR INEXIT	;CHECK FOR CR ;EXIT IF YES
9265 9266	F0 FF30		LDX SMI	838H	RE-GET CHAR
8268 826A	385E FF 0 A		em SMI	eah	ELSE TEST FOR
026C 026E	3878		BM	DECIML	DECIMAL, GO CONVERT IF YES
8270 8270	3B5E FFAA		BM SMI	ر ILOOP	;TRY AGAIN IF YES :ELSE TEST >= #46
8274	335E		BPZ	IL00P	TRY AGAIN IF YES
0276 0278 027A	FC96 FC9A AF	DECIML:	adi Adi Plo	6 0AH RF	;else ok hex a-f ;make 4 bit value ;save temp

TREC

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PROGRAM

627B	347B	INLOOP:	B1 offfe	INLOOP	WAIT FOR TRE
0270 0275	63 12			P 2	INDA ANTAINCR
0272	22		DEC	112	JOINTO HOLDINON
		;SHIFT	rcvd 4 B	it value into "I	IODAT" FROM R TO L
027F	99		GHI	R9	;GET BUFFER HB
0280	FE		SHL		
6281	FE		SHL		
8282	FE		Shl		
8283	FE		SHL		;SHIFT L 4
0284	52		STR	R2	;put on stack
8285	89		GLO	R9	;BUFFER LB
8286	F6		SHR		
0287	F6		SHR		
8288	Fó		SHR		
6289	F6		SHR		;SHIFT R 4
828A	Fi		OR		;MERGE HBLN+LBHN
828B	89		PHI	R9	, SAVE AS NEW HB
928C	89		GLO	R9	;BUFFER LB
828D	FE		SHL		
028E	FE		SHL		
628F	FE		SHL		
6290	FE		SHL		;SHIFT L 4
0291	52		STR	R2	;put on stack
8292	8F		GLO	RF	RE-GET NEW NIBBLE
8293	Fi		ŨR		MERGE LBHN+NEUNIB
0294	A9		PL0	R9	SAVE AS NEW LB
8295	3 0 5E		BR	IL00 P	GET NEXT DIGIT
0000			END		

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TREC	
	A1.A.

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	SYMBOL	TABLE	

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	882E	DECIM	8278	11.0 0P	825E	NOTILT	8881	PLAYBK	8888
	8888	DELCOP	6692	INDAT	825A	OCEXIT	00DD	PROMPT	8 15D
	882E	DLAY	8874	INEXIT	8259	OCLOOP	00E2	RESTRT	88A2
	0899	Dlaymk	8948	INIT	0131	ONEXIT	0045	RSLOOP	88A5
ACLOOP	8244	DTILT	809E	INLOOP	027B	ONLOOP	084E	SP	8828
AQLOOP	8906	DYLOOP	887C	LF	998A	OTEXIT	002E	STKTOP	8199
AQUI	00B4	ENDCHK	80F6	LINSUM	828F	OUTCHR	00DE	TSEC	80E9
BURN	9 IAD	ETX	8083	MAIN	9854	outdat	802F	TSEXIT	80E8
Capgee	6947	GDEXIT	80D4	Movdat	8614	OUTNUM	8846	TSLOOP	88EC
CKLOOP	0215	GETDAT	99D5	NOIDLE	886C	Param	6181	VERIFY	0 1B9
CR	888D	6LITCH	007B	NOINST	09C4	PBEND	082 3	VLOOP	81C 9

|--|

	•• •					• • • • • • • • • •	- 7
F1_	Loc	COSMAC	CODE	LN	NO	SOURCE LINE	-
	6666				1	PLAYBACK ROUTINE FOR DEDR TEMP RECORDER	
	8888				Ē	FOR STANDALONE USE	
	3000				. 3	IN EMERGENCY STILLATIONS	~~.
					"	T E WHEN "TREC" PLAYBACK POUTTNE EATLE	
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	Ch l'h l'h l'h					TUYE BEARDAM FAN DE LAADED EDOM	
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	남남남남				1.		<u> </u>
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	8988				14	***********************************	米米赤米
	0000				15	••	-
	លូលលូក				14	PROGRAM COUNTERS	
	9990				17	••	
	0000				18	MAINPC = #00 MAINLINE ROUTINES PC	~
	장갑감장				17	TSECPC = 401 "TSEC" PC	
	ចលិចដ		. •		已边	OUTDPC = #04"OUTDAT" PC	
	0000			•	21	OUTCPC = #06OUTCHR" PC	-
	0000				22	OUTNPE = #03 "OUTNUM"PC	
	0000				23		
	9999				24	POINTERS AND FLAGS	-
	2000		•	1	25		
	0000			,	26	DATPTR = #07DATA MEMORY POINTER	
	3000			;	27	STKPTR = #02STACK POINTER	
	9000				28	ENDFLG = #0D END-OF-DATA-AREA-FLAG	
	0000			I	29		
	****				30	BUFFERS AND WORKING STORAGE	
	00000				31		
	0000				32	IODAT = 409DATA FROM/TO I/O ROUTINES	
	0000				33		
	0000				34	, COUNTERS	
	លាលាលា				35	•••	•
	8889				Зó	TSCTR = #0B "TSEC" INTERVAL COUNTER	-
	0000				Э7	• •	
	4444				38	SERATEHPAD	
	0000				34	· · '	
	0000				40	SCRA = #0FSCRATCHPAB REGISTER	
	0000				41	• •	
	0000				4 E	• •	-
	0000			· ·	43	••	
	0000				44	*************** EQUATES AND CONSTANTS *	本本本本
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	0000			4	4.6	I/O PORT NAMES	
	0000			4	47	• •	
	0060	•			43	UARTOT = #05 UART XMITTER BUFFER REGIS	75-9
	0000				47	POWRDN = #07POWER DOWN	
	59999				50	••	
	0000				51	CONSTANTS	
	0000				52	••	
	아이라지				53	LF = #ØALINEFEED CHARACTER	
	<u>3000</u>			1	54	CR = #0DCARRIAGE RETURN	
	0099				55	SP = #20SPACE	
	0000			1	56	ETK = #Ø3´END-OF-TEKT CHARACTER	
	0000			1	57		~
	0000			1	58	· ·	
	0000			1	99	···***********************************	法教育法
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	0999	,			5 2	· .	
	0000				63	· · ·	~
	0000				<u>6</u> 4	INITIALIZE WORKING REGISTERS AS REG/D	

ساق ا	<u>.</u>	ីថាដា						······································
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	۲ ۳ .	<i>11 11 11 11 11</i>	(* (34/14)	00		1 <i>D</i> .1.	A.1(15EL)	INITIALIZE
		0002	B 1.	47		PHI	TSECPC	TSEC PC.
	F	0003	F800	48		LDI	A.Ø(TSEC)	
- '		AAAC	A 4			5.5	THEFT	
			μ. T			(~ i i./	(DELFL	
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	Æ	0004	F800	71		LDI	A.1(OUTDAT)	"OUTDAT" Pr
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	ş.,	<i></i>	F 3 0 0	/5		1-1) L	A.I.OUTCHR)	"OUTCHR" PC
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		19491,1	A4	14	•	8L0	- UU ILPL	
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	F	លលារទ	F800	83		LDI	A.Ø(OUTNUM)	
		0017	AB	84		8L.0	OUTNPC	
		CARLE C		oc				
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	÷۳	<i>01</i> 918	F 3 40 M	85		LDI	A.1(STETOP)	STACK POINTER
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0046	133.	133		SEF ISECFL	
0047		1,34			
0047	3028	1.35		BR PLAYEN	KEEP SENDING DATA
9947		135			
ማዝዋል	F) A	137	SBEMD:	SEP NUTEPE	SEND MSG WHEN THR
WARA A	EDEAACEG	3 3 0			
427 927 AP (A)		4.20			
1919 A E	# 3# D Y A # #	3.37		/ETX/LR/LF/#WW	
9 9 5 2		140	• •		
005 2	óF	141		INF FOWFON	THEN POWER DOWN 🦱
0053	动动	142		IDLE	. AND IDLE FOR SAFET
005 d		143			· · · · · · · · · · · · · · · · · · ·
and a		4 8 8	••		
1.1 "Li Li -14 1"Li "Li -14		4 4 M	· ·		the with a state that a state of a standard and a state of a state
8054		145		k nederik nederik nederik nederik (1777-177	WINDING RARAWAR
잡습도 수		146	• •		
G 25 4		147	"OUTDAT'	' TRANSMITS CONTENT	IS OF REGISTER "IOPAT
6665-4		148	"OUTDAT"	' ENTERED VIA "SEP	OUTDPC" AND EXITS VI
6654		149	\'		
(B) (B) (B)	T1 (3)	160	orstvrr:	SED MATNEC	PETHON TO CALLEP
~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~	10 M	1	547 S Ban 2 S pl. 3 -	µriuni 110°1,461%1 ∿un '	to built of the test of the first sectors in the sector of the secto
12.62.5		151			
0055	÷÷	15 <i>E</i>	OUTDAT:	GHI IQBAT	HIGH BYTE TO XMIT
005S	FAFAFAFA	153		SHR; SHR; SHR; SHR	ISOLATE HIGH NIPPL
9005 A	БЭ	154		SEP OUTNPC	SEND IT
005 R		155	· · · ·		
(A (A C U)		164		GUT TORAT	UTCH BYTE
*******	1			CHELL L. 4./14/9911 A 8/27 AA/22/27	
194951	r avar	157		AN1. 4497	LUW NIBELE
995 E	EG	158		SEP OUTNPC	
905 F		157			~
2/28 F	89	160		GLO IODAT	LOW BYTE
<u>ខាង៩ល</u>	FAFAFAFA	isi		SHR; SHR; SHR; SHR	HIGH MTRREF
3346	n 9			CCO AUTHOR	
014103-44 (M.M. C. M.	13 5	1.000		SEE GOINES	
- 0065		193			
0065	87	154		GLO IODAT	LOW BYTE
8966	FAOF	145		ANI #ØF	LOW NIBBLE 🥱
0063	ÐЭ	166		SEP OUTNPC	
MMA9		167	•		
0-0-1 D	15 64 FL 4	4	••	DD DTEVIT	DECET NONTROPY DEE
00007	15 65 25 44	100		DR DIEVII	HEDE: DOIDFE FAR
0066B		167	• •	-	
00-0-5 B		174		- DEDICATED SOBE	CONTINE FUR "UDIDAT".
边的古田		171	VALUE IN	I D REG TO ASCII HE	EX (0-9, A-F) AND MMI
0045		172			
0048	T) 4.	173	ONEXIT:	SEP OUTDPC	RETN TO "OUTDAT"
174 IA 3. 17		174			
10 10 C L	cute (par per, ),	.L. / ""	· ·	··· • • • • • • • • • • • • • • • • • •	······································
WWAL	rrøa	1/5	00440933	SMI AGA	$\dots (ES) FOR (= 7)$
@@4E	C /	176		LBPHP	SELF ADD IF YES
996F	FCØ7	177		ADI #97	IF A-F, ADD EXTEA
9071		173			
997i	FC 34	179		ADI NJA	ADD NET 30 OR
		190			MET BA TO GET
4343 C 3 15.19.19		a. 6. 4.) 4. 65. 4			
667 s		767		Jan augo guna,	. MOLLL LPMMMALIER.
0073	in m	182		SIN SIRFIN	FUI UN STACKI NU P
9974	3474	133	ONLOOP:	B1 ONLOOF	WAIT FOR UART TPE
0076	<u>45</u>	184		OUT UARTOT	SEND TO UART
aa77	22	185		DEC STRPTS	UNDO AUTOINCREMENT
	764.0	1 3 4		BD OMENTT	GENET CONTRACT
4747733 1921-1923	aat food haad	3 0 0 3 0 0		warren in the second for the fi	a a second constant and the first to the terms of the second constant of the second constan
1010 / A		10/	، ، • •	and a standard and a	1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1
9997A		1.3535	ં સંગળવામાં સંગળવામાં સંગળવા	n she	late a substant and a state of a state of the state of th
007A		189			
007A		190	EN7	FERED VIA "SEP OUTC	PC". OUTPUTS CHARAC
<b>位码了</b> 本		191	. POINTED	TO BY "MAINPC" UNT	TIL A 00 BYTE IS ENCO
(А.В.Т.А.		100	RETHRMS	VIA "SEP MAINPE"	ALTERS MISTEPTR)
4/4///********************************		400	· · · · · · · · · · · · · · · · · · ·	som ti tar en st tijt≕tide 1361 'ns e	·
10107A	<b>771</b> .411	1.7 m 1.7 m	չուլ չուցող է չեղել դրուլ։ Հե		SCTRON TA ANYES
10 10 / A	1.7.49	174	Q4.25.1 1	SET MALNEL	
ØØ78	417	195	OUTCHR:	LDA MAINFC	.GEN CHARACTER
007C	327A	196		BZ OCEXIT	IF #00, EXIT

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··.

	007E				- STR STKPTR	THE FIT OF ACL
	007F	, 347F	198	OCLOOP:	B1 OCLOOP	. WAIT FOR MART O
	0081	. 65 .	199		OUT UARTOT	. XMIT WHEN READY
	008a	22	200		DEC STKPTR	UNDO AUTOINCR.
	0683	307B	201		BR OUTCHR	. GET NEXT CHAR
	0085	•	202			
	0085	,	203	, 宋字宋永宋末月	********	SEC *************
	0085		204			0
	0085		205	"TSEC"	IS A 0.1 SEC	TIMING ROUTINE. TIME FROM
	0035		206	. EXECUTI	CON OF FIRST I	NSTRUCTION (DEC TSCTR) TO
<u> </u>	0085		207	OF LAST	F INSTRUCTION	(SEP MAINPE) IS
	0085		209	(2	D.1) X (VALUE	IN "TECTR") SECONDE
	ØØ85		209	. ASSUMIN	46 A 38.400 KH	Z CPU CLOCK.
-	0065		210	. CALL MI	TH "SEP TSECF	C". RETURNS TO MATNER 496
	0085		211			
	<i>98</i> 35	E) (2)	212	TSEXIT:	SEP MAINPC	. RETURN TO FALLER
<u> </u>	0085	28	213	TSEC:	DEC TSCTR	DEC COUNTER -
	0087	F874	E14		LDI #74	SET UP LOCAL LOUP
	0089		215			
~	6689	FFØ1	216	TSLOOP:	SMI #01	-
	008B	3A89	Z17		BNZ TSLOOP	
	0080		218			
-	003D	9 B	514		GHI TSCTR	. CHECK COUNTER -
	008E	5 <i>2</i>	220	,	STR STKPTR	. PUT HE ON STACK
	008F	88	221		GLO TSCTR	GET LB
	0090	F1	222		OR	MERGE HB+LB 🔷
•	0071	2225	223		BZ TSEXIT	EXIT IF BOTH Ø
	0073		224	• •		
	10093 E9001	3086	ees		BR TSEC	ELSE KEEP LOOPING
	ØØ95		226	· · · · · · · · · · · · · · · · · · ·	લ્લા એટલે ગો છે છે. તે ગો છે છે છે છે.	DATA AREA 米非市中非非市中市中
	0095 5		287			
•••	Ø180		.223		ORG #100	
	(9 1 9 9) (9 1 9 9)		229	STKTOP:		
	9191		230 065		ORG #+1	
	0101		231	Радат:		
	0101		232	END		
	NO UNDE	FINED NAMES				· · · · · ·

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c. ADTEST

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ADDRESS CODE MNEMONIC COMMENT

			NAME OF PROGRAM, ADTEST
			_} ************************************
0000	F8	LDI	;MAKE REG 3 = 00C8
0001	C8	C8	;MAKE REG THE ADDRESS OF X OUT IN THE BOONIES
			WHERE IT WON'T HURT ANYTHING
0002	A3	PLO	PUT LOW REG 3
0003	F8	LDI	
0004	00	00	
0005	R3	PHI	PUT HI REG 3
0003	E3	SEX	TO REG 3
0007	78 78	SET D	TURN ON AZD POWER
			********
			START A/D CYCLE AGAIN
0009	ZÞ	THP 9	TDIGGED A/D
0000	00	1141 0	LIAIT LOOD COD MEAGUDEMENT DEDETITION
			WHIT LOOF FOR MEHBOREMENT REFETITION
0000	50	1.07	j RHI L
0009	F8		
000A	++	44	;WAIT TIME
0008	FF	SMI	;SUBTRACT MEMORY IMMEDIATE
000C	01	01	;NUMBER SUBTRACTED
000D	64	NOP	;WASTE TIME
000E	, C4	NOP	
000F	C4	NOP	
0010	C4	NOP	
0011	C4	NOP	
0012	C4	NOP	
0013	<u>64</u>	NOP	
0014	C4	NOP	
0015	30	BN7	SHORT BRANCH IE NOT 0
0012	ົດອ		RRANCH LOCATION
0010	00	00	* NARAZANAN CONTINUS.
			PEAN A/N AND SEND IT OUT UTA HART
0017	9F	ON O	CUODE BOANCH IS SEND IN OUT TE AVE STATUS IS
0017	JC	ONO	(SHUK) BRHNUN IF EFS-U, WHIT IF HZD STHIUS IS
0040		4 7	
0018	17	17	BRANCH LUCATION
0019	67	INP 1	;INPUT 1, HI URDER BYTE FROM A/D
001A	65	007 5	; OUTPUT 5, OUTPUT HI URDER BYTE TO DART
0018	23	DEC	DECREMENT REG 3, CANCELS AUTO INCREMENT IN
	<i></i>	<b>-</b>	REG 3 DUE IN UNI 5
0010	34	81	SHURI BRANCH IF EFIFI, WAIT TIL UART READT
0010	1	10	BRANCH LUCATION
001E	6A	INP 2	(INPUT 2, LU URDER A/D BYTE
001F	65	OUT 5	OUTPUT 5, OUTPUT LO ORDER BYTE IN WARI
0020	23	DEC	;DECREMENT REG 3 (X REG), CANCELS AUTO
			;INCREMENT DUE TO OUT 5
0021	34	B1	;SHORT BRANCH IF EF1=1, WAIT FOR WART TO
			;FINISH
0022	21	21	;BRANCH LOCATION
			; * * * * * * * * * * * * * * * * * * *
			END OF READING OUT A/D
0023	30	BR	SHORT BRANCH. START AGAIN
0024	08	08	BRANCH LOCATION
~~~ '	* *		,

d.	MEMTE	ST2		
ADDR	CODE	LABEL	MNEMONIC	COMMENT PROGRAM NAME - MEMISI2
0000 0001 0002 0003 0005	C4 C4 78 F8AA A4		NOP NOP SEQ LDI #AA PLO R4	;TURN OFF A/D CONV ;LOAD IN TEST PATTERN ;SAVE IT IN R4.0 ;************************************
0006 0008 0009 000A 0008 0008	F800 B3 A3 B2 F860 A2	START	LDI #00 PHI R3 PLO R3 PHI R2 LDE #60 PLO R2	;SET UP OUTPUT POINTER ;SET UP MEMORY TEST POINTER ;************************************
000E 000F 0010 0011 0012 0014	84 52 12 92 FFOC 3BOE	FILL	GLO R4 STR R2 INC R2 GHI R2 SMI #0C BM FILL	;FILL MEMORY WITH TEST PATTERN ;LOAD TEST WORD ;PUT IN MEMORY ;POINT UP TO NEXT LOCATION ;GET MEMORY PTS ;TEST FOR DONE ;IF NO, KEEP FILLING
0016 0018 0019 0018	F800 B2 F860 A2		LDI #00 PHI R2 LDI #60 PLO R2	RESET R2 TO BEGINNING OF TEST AREA
001C 001D 001E 001F 0021 0022 0023 0025 0027 0029 0029	E2 84 F3 3A35 12 92 FFOC 3B1D F847 53 E3	TEST	SEX R2 GLO R4 XOR BNZ NOTOK INC R2 GHI R2 SMI #0C BM TEST LDI #47 STR R3 SEX R3	;COMPARE MEMORY WITH TEST PATTERN ;SET R2=X ;GET TEST PATTERN FROM R4.0 ;COMPARE TEST PATN & MEM CONTENT ;IF "NOT OK," BRANCH ;ELSE INCREMENT MEM POINTER ;GET POINTER ;TEST FOR DONE ;KEEP TESTING IF NOT DONE ;LOAD "G"
002B 002D 002E 002F 0030 0032 0033	3428 65 23 84 FBFF A4 3006	GLOOP	B1 GLOOP OUT 5 DEC R3 GLO R4 XRI #FF PLO R4 BR START	WAIT FOR WART FINISH OUTPUT TO WART UNDO AUTO-INCREMENT GET TEST PATTERN COMPLEMENT IT STORE NEW TEST BYTE REPEAT TEST MEMORY ADDRESS AND ERROR OUTPUT ROUTINE
0035 0036 0037 0039	E3 53 3437 65	NOTOK BLOOP	SEX R3 STR R3 B1 BLOOP OUT 5	; ;POINT TO OUTPUT REG ;STORE ERROR WORD IN OUTPUT REG ;WAIT UNTIL UART READY ;UART OUTPUT

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3. MACHINE CODE FOR TEMPERATURE RECORDER (xxx.ASC FILES)

a. TREC.ASC

1.125-11104	0 F F	VUGI	U-8-1														
0000 : CO	01	31	E2	98	B7	88	A7		0188:5	jΕ	F8	00	B3	F8	46	A3	F8
0008:97	52	9D	F7	·ЗА	14	87	52		01C0:0	1	88	F8	01	A8	D6	0A	ÛΑ
0010:8D	F7	32	23	47	B9	47	A9		0108:0	0	D6	ÛA	0D	80	F8	02	A8
0018:04	D6	00	ΟA	00	F8	02	A8		01D0:D)1 -	48	89	52	48	A9	F1	C2
0020:D1	30	80	D6	53	54	4F	5û		0108:0	2	0F	D6	23	53	41	4D	50
0028:03	0D	ΰA	80	6F	00	DO	99		01E0:4	IC -	45	53	3D	20	00	D4	48
0030:F5	56	56	F6	D3	99	FA	OF		01E8;E	39	48	A9	06	20	20	20	49
0038:D3	89	F6	F6	F6	F6	D3	89		01F0:4	E	54	45	52	56	41	40	3D
0040:FA	0F	D3	30	2F	D4	FF	ñΑ		0158:2	20	00	<u>n4</u>	48	89	48	Δ9	50
0048:07	FC	07	FC	34	52	34	4F		0200:2	20	20	20	44	45	40	41	59
0050:65	22	30	45	48	85	52	48		0208+3	RD -	20	00	54	rn.	л ў	rg.	FR
0058.40	F1	20	.с. сс	RF	ΔF	FR	n۵.		020010	in I	R7	Δ7	pq.	57	89	F4	20
0060.55	15	SF	15	55	97	RD	87		0210.0	7	97		00	-r -	02	17	27
0060.00	70	5	00	10	 D-0	10	λQ		021011		57	n D	00	15	52	10	20
0000.00		0F //0	~~	90 GA	02	40	ND ND		022011	с. 26 -	20	40	02 40	10	40	10	20 50
0070:40	DH	40	HH FO	211	ロレ	OH OC	HL Fi		022012	:U : :E	20 40	43	98	93	93	40 NC	03
00/8:18	36	HE OF	52 00	30	52	20	11		0230:0) 0 - C(40	30	20	00	1/4	06	20
0080:32	84	38	32	AZ	6E	18	40		023812	20	20	41	43	54	49	41	4Ł
0088:52	65	22	20	18	09	AB	01		0240:3	}}`	20	03	00	35	44	6C	FF
0090:F8	49	FF	01	7A	3A	92	3F		0248:4	17	CA	91	31	98	87	88	A7
0098:9E	F8	30	AE	30	70	2E	7A		0250:F	8	01	88	F8	01	A8	CO	90
00A0:30	7C	F8	4B	AE	8E	32	74		0258:5	54	DQ	F8	00	B9	A9	35	5E
00A8:F8	28	AB	D1	37	Bĩ	F8	4C	,	0260:6	50	FF	0D	32	59	FO	FF	30
0080 :AE	2E	30	A5	9D	52	8D	F1		0268:3	38	5Ē	FF	0A	ЗB	78	FF	07
0088:32	54	78	6E	2D	99	8B	89		0270:3	3B	5E	FF	06	33	5E	FC	06
00C0:AB	2B	D1	68	F8	6A	FF	01		0278;F	Ç	θA	AF	34	78	65	22	99
00C8:3A	СS	D5	97	FF	0C	38	84		0280:F	E	FE	FE	FË	52	89	F6	F6
0000:27	27	30	54	DO	69	57	17		0288:F	6	F6	F1	89	89	FE	FE	FE
00D8:6A	57	17	30	04	DO	40	32		0290:F	E	52	35	-i	A9	30	5E	FF
00E0:0D	52	34	E2	65	22	30	DE		0298 : F	F	FF	FF	FF	FF	FF	FF	ZZ
00E8:D0	28	F8	74	FF	01	3A	EC										
00F0:9B	52	88	F1	32	E8	30	E9										
0058:55	FF	FF	FF	FF	FF	FF	FF										
0100:EE	FF	FF	FF	FF.	FF	FF	FF										
0108-FF	77	FF	FF	- - -	FF	FF	FF										
0110+55	FF	33		न्न	FE	FF	37										
0118.FF	FF	FF	EE.	EE.	FF	FF	11										
0110171	55	55	55	EE.	E E	55	55										
012011			CC.			EE	CC.										
0120170	76	77 20	55	ГГ 01	TT TO	E CE	ГГ А1										
0100.00	02	го 00	00 CO	EV DT	70 70	22	NU HI										
0140-04	02	20	ГО АЛ	3H 70	HJ AA	F O DC	00										
0140:04		25	H4	rö ne	00	DD											
0148:01	Hb Ab	10	00	80	10	50	HO.										
0100118	01	82	18	00	RZ RC	18	01										
0158:88	18	01	A8	£2	D6	UA	UD										
0160:00	F8	00	BB	£8	02	AB	D1										
U168:D6	23	53	41	4D	50	40	45										
0170:53	ЗF	20	03	00	D3	99	58										
0178:52	18	89	58	18	F1	32	AD										
0139:06	20	20	20	49	4E	54	45										
0188:52	56	41	40	3F	20	83	00										
0190:03	99	58	18	89	58	18	D6										
0198:20	20	20	44	45	4C	41	59										
01A0:3F	20	03	00	D3	99	58	18										
01A8:89	58	18	30	5D	F8	00	8F										
01E0:AF	1F	F8	00	5F	1F	F8	7B										

,

b. AVERG.ASC

AVERG ASC PROGRAM

0000:00	01	A2	Ε2	98	87	88	A7	0188:F	8	4F	A6	F8	01	85	F8	15
0008:97	52	9D	F7	ЗA	14	87	52	0100:4	15	F8	01	B2	F8	71	A2	F8
0010:8D	F7	32	23	47	89	47	A9	0108:0)1	B8	F8	72	8 8	Ε2	D6	0A
0018:D4	D6	0D	0A	00	F8	02	AB	01D0:0	Ð	00	F8	00	BB	F8	02	AB
8020:D1	30	08	D6	53	54	4F	50	0108:0)1	D6	23	53	41	4D	50	4C
0028:03	0D	ÛΆ	80	6F	00	DO	99	01E0:4	5	53	3F	20	03	00	D3	99
0030:F6	F6	F6	F6	D3	<u>99</u>	FA	0F	01E8:5	58	52	18	89	58	18	F1	C2
0038:D3	89	F6	F6	F6	F6	03	89	01F0:0	12	20	06	20	20	20	49	4E
0040:FA	OF	03	30	2E	D4	FF	ΰA	01F8:5	4	45	52	56	41	40	3F	20
0048:07	FC	£7	FC	34	52	34	4F	0200:0	13	nn	03	99	58	18	89	58
0050:65	22	30	45	48	RD	52	48	0208:1	8	ກຄ	20	20	20	44	45	40
A058:AD	F1	34	60	BF	ΔF	FR	C4	A210:4	11	59	3F	20	03	ກົດ	D3	99
0060:55	1F	5F	1F	5F	97	8D	87	0218:5	8	18	89	58	18	C0	£1	ĈE.
0068,40	76	GF	00	48	89	48	Δ <u>9</u>	0220.4	ΞQ.	00	BF	ΔF	16	FR	nn.	SE
0070+49	PA	49	20	96	BC	96		022011	5	FQ.	78	55	τQ	00	Q2	50
0070-59	20	ΔF	F2	9r	52	gr	Fi	022011	נג וב	23	FR	01	20	EB	72	
0020.22	R4	QE	32	Δ <u>2</u>	25	FQ	40	0200.7	10	ñΔ	00	00	50	10 ΩΔ	00	но 60
0098.52	65	22	20	FQ	09	ΔR	Di	020010	9	0H 02	ΔR	D1	48	pq.	52	48
0000.52	49 49	FF	01	70	20	92	SE	0249+4	a.	51	02	02	22	DD ΛC	22	52
0020.00	50 50	20	VE.	20	70	25	70	025010	1_7 11	4D	50	102	45	50	20	20
00100100	70	ас ГО			00	25	711	023014	10 10	57 64	40	70	7J 10	73 VØ	50	20
0050.50 00A0.E0	20	TU AD	D1	⊡⊑ 97	01	ED.	7 T 4 C	020010)))	20	49	AE	5/	n√ 15	50	50
00000.00	20	20	Λ <u>Ξ</u>	37 66	01	00	70 AD	0200:2	.V 14	20 70	97.7 OD	9C 20	00	40 54	10	30 DQ
0020185	70	ос ап	70 50	99 010	00	20	110 121	020014	11 10	70 AQ	00	20	20	20	70 //	D2 AE
0000370	7 D 0 0	DA	75	ου το	50	32	J4 25	027014	10 10	H7 41	50	20	20	20 00	74 64	90 60
0000110	00	0H 20	AH QD	70 50	00	J <u>2</u> 51	0J 03	02/0:4	10 10	9C	52	00	20	00	09	57
0000122	00 50	2D 1 5	20	JZ 01	20	LT 1	20	020010	2	36 57	го 70	00 47	<u>זם</u> רמ	EC.	07 01	57 CD
0000100	со 05	19	7.7 00	UI.	00	20	20	020010	2	777 00	82 07	17	21	f F CD	01 07	00
0000:05	2H 5A	CH QA	00 50	ГD СС		7.9 0A	CD	023010	2	00 62	20	7 F 17 A	20	10	02 40	00 45
0020:20	<u>2</u> H ^^	211	го 52	CA	CHI CA	CA	ГD DA	022010	2	10	50	20	20 4D	נויי חרי	40 20	9J 00
0050177	13M 70	10	Π.C. GA	57	ст сс	50	C7	0200.0	io M	70 NC	20	20	20	20 //1	40 40	50 5.4
00F0:75	17	75	56	ro EC	60 56	57	GV LT	028010	10		20	20	20	0.2 4T	10	25
00100.57	т. СС	CC.	E E	50	F0	57	 17	020017	2 7	ግ። ሬድ	55	17	CA	03	Δ <u>0</u>	00 00
0100375	67		гс. 60	FE.	Γ.L.	07 DA	27	0200:0	77 17	00	ΓΓ Λ7	77	01	DO DT	HC CO	20 70
0100120	27	εr 00	54	00	EO.	00	27 AC	020010	0	00	H7 00	F0	00	00 50	60 00	12
0110:27	22	100	74 20	CD	70 70	UO EO	NC OC	020018	10 .0	25	00	04 60	DO EE	10 10	00 22	02
0100.55	02 03	19	20	20	70 EA	ГФ. 00	00	02001H	12	3J 20	00	00	ГГ 04	00	36 08	00 00
0120.77	CA .	OA	20	07	CH CR	00 EA	02 05	020017		55 55	07	00 00	D1	EE.	UH DC	3D 3D
0128:38	64	800 74	174 54	AM TO	20	rn JD	10	022032	B	rr co	07	35	01	rr	00	33
0130:32	28	74	EA CO	78	78	78	/8 50	02531	л .е	11	00	76 77	VA FF	Ar	39 FF	50
0138:30	18	Ho AA	18	55	10	01	52	0250:6	50	66.	33	FE.	ft cc	FE.	FE.	02 02
0140184	74	AA TA	63	18	10	78	11	0218:2	53	76	10	10	10	11	89	89
0148:52	9A	79	БĤ	30	18	00	40	030011	۲ <u>۲</u>	FE PO	11	12	32 55	18	11	A9 EE
0100:32	42	32	34	23	60	22	30	0308:0	.U	02	D1	77	11	۲۲	51	77
U158:4F	00	ZB .	18	74	hh co	U1 50	38	0310:0	10	~						
0160:50	38	52	88	11	32	22	30									
0168:5A	11	11	11	} F F	FF	11	++									
01/01F	11	F7	1	}}	F}	}}	ነት ተታ									
01/8;FF	FF	łF	FF	۶F	FF	FF	FF									
0180:FF	ł F	FF	FF	FF	łF	FF	11									
0188:FF	۲F	FF.	++	FE	14	ነት የተ	FF									
01901FF	ጉተ 	27 	77	۲۲ 	11		77									
0198:FF	łF	FF	FF	۶F	FF	١F	FF									
01A0:FF	FF	7A	F8	01	81	F8	5A									

01A8:A1 F8 02 B3 F8 CD A3 F8 01B0:00 B4 F8 2F A4 F8 01 B6 .

.

c. PLYBK.ASC

T

PLYBK ASC PROGRAM 0000:F8 00 B1 F8 86 A1 F8 00 0008:84 F8 55 A4 F8 00 B6 F8 0010:78 A6 F8 00 B3 F8 6C A3 0018:F8 01 B2 F8 00 A2 E2 F8 0020:08 BD F8 FF AD F8 01 B7 0028:F8 01 A7 97 52 9D F7 3A 0030:37 87 52 8D F7 32 49 47 0038:B9 47 A9 D4 D6 0D 0A 00 0040:F8 00 BB F8 02 AB D1 30 0048:28 D6 53 54 4F 50 03 0D 0050:0A 00 6F 00 D0 99 F6 F6 0058:F6 F6 D3 99 FA 0F D3 89 0060:F6 F6 F6 F6 D3 89 FA OF 0068:D3 30 54 D4 FF 0A C7 FC 0070:07 FC 3A 52 34 74 65 22 0078:30 6B D0 40 32 7A 52 34 0080:7F 65 22 30 7B D0 28 F8 0088:74 FF 01 3A 89 9B 52 8B 0090:F1 32 85 30 86 FF FF FF 0098:FF FF FF FF FF FF FF FF 00A0:ZZ

d, ADTEST.ASC

ADTEST.ASC PROGRAM 0000:F8 C8 A3 F8 00 B3 E3 7B 0008:6B F8 A8 FF 01 C4 C4 C4 0010:C4 C4 C4 C4 C4 C4 A 0B 3E 0018:17 69 65 23 34 1C 6A 65 0020:23 34 21 30 08 ZZ

e. MEMTS2.ASC

MEMTST2.ASC PROGRAM

0000:C4	64	7B	F8	ΑA	Α4	F8	00	
0008:B3	A3	B2	F8	60	A2	84	52	
9010:12	92	FF	00	3B	0E	F8	Û8	
0018:82	F8	60	A2	Ε2	84	F3	3A	
0020:35	12	92	FF	0C	3B	1D	F8	
0028:47	53	E3	34	28	65	23	84	
0030:FB	FF	Α4	30	06	Ε3	53	34	
0038:37	65	23	92	53	34	3D	65	
0040:23	82	53	34	43	65	23	F8	
0048:00	53	34	4A	65	23	E2	30	
0050:21	04	22						

APPENDIX G. PRO-350 PROGRAMS

1. PROBASIC.BAS

SET NO DOUBLE

5 PROGRAM PROBASIC 10 PRINT "FOR HEATFLOW TYPE 'RUN MENU' (CR)"

. 2. MENU.BAS

SET NO DOUBLE

5 PROGRAM MENU

10 HOME\$=CHR\$(27)+*[2J*+CHR\$(27)+*[0;0H*\INVERSE\$=CHR\$(27)+*[7m*

15 FLASH\$=CHR\$(27)+*[5m*\NORMAL\$=CHR\$(27)+*[0m*\CENTER\$=CHR\$(27)+*[12;30H*

20 PRINT HOME\$\ PRINT TAB(38); INVERSE\$; "MENU"; NORMAL\$\ PRINT \ PRINT

30 PRINT "1. THERMAL CONDUCTIVITY OPERATIONS"

40 PRINT "2. DOWN HOLE INSTRUMENT OPERATIONS"

99 PRINT *9. EXIT BASIC*

100 PRINT \ PRINT FLASH\$;"INPUT YOUR SELECTION ";NORMAL\$;\ INPUT A\$

110 IF A\$<"1" OR A\$>"9" THEN GOTO 20 ELSE A=VAL(A\$)

120 CN A GOTD 140,150,20,20,20,20,20,20,220

130 GOTO 20

140 PRINT HOME\$;CENTER\$;INVERSE\$;*GETTING THERMCON PROGRAM*;NORMAL\$

145 CHAIN "THERMCON.BAS"

150 PRINT HOME\$;CENTER\$;INVERSE\$;"GETTING LOADER PROGRAM";NORMAL\$

155 CHAIN "LOADER.BAS"

220 END

3. LOADER.BAS

```
SET NO DOUBLE
5 PROGRAM LOADER
19 REM DSDP INSTRUMENT PROGRAM MODIFIED FOR PRO-350
11 REM BY GEORGE L. PELLETIER WHOI 19 DEC 85
20 HOME$=CHR$(27)+*[2J*+CHR$(27)+*[0:0H*\INVERSE$=CHR$(27)+*[7m*]
25 NORMAL$=CHR$(27)+"[Om"\FLASH$=CHR$(27)+"[5m"\CENTER$=CHR$(27)+"[12;30H"
30 PRINT HOME$+INVERSE$+** CORING TOOL TEMPERATURE RECORDER **+NORMAL$\ PRINT
35 PRINT "MENU"\ PRINT "1. LOAD PROGRAM"\ PRINT "2. LOAD PARAMETERS"
40 PRINT *3. DUMP DATA*\ PRINT *4. DISPLAY FILE DATA*
45 PRINT "5. PRINT FILE DATA"
50 PRINT "6. CONVERT & PRINT FILE DATA"\ PRINT "7. END"
100 PRINT FLASH$;"ENTER YOUR CHOICE ";NORMAL$;\ INPUT A$\A=VAL(A$)
110 ON A GOTO 120,1000,2000,3000,4000,5000,6000
115 GOTO 30
120 PRINT HOME$+INVERSE$+"*** LOAD PROGRAM ***"+NORMAL$\ PRINT
125 PRINT "1. LOAD MEMORY TEST"\ PRINT "2. LOAD MAIN PROGRAM"
130 PRINT "3. LOAD A/D TEST"\ PRINT "4. LOAD AVERAGING PROGRAM"
135 PRINT "5. LOAD PLAYBACK PROGRAM"
150 PRINT FLASH$;"ENTER YOUR CHOICE ";NORMAL$;\ INPUT A$\A=VAL(A$)
155 IF A(1 OR A)5 THEN 30
160 IF A$="1" THEN PG$="MEMTS2.ASC"\ GOTO 250
170 IF A$="2" THEN PG$="TREC2.ASC"\ GDTO 250
180 IF A$="3" THEN PG$="ADTEST.ASC"\ GOTO 250
190 IF A$="4" THEN PG$="AVERG.ASC"\ GOTO 250
200 IF A#="5" THEN PG#="PLYBK.ASC"\ GOTO 250
250 OPEN PG$ FOR INPUT AS FILE #11/ REM GET READY TO LOAD PROGRAM
255 PRINT \ PRINT INVERSE$;"SWITCH INSTRUMENT FROM RUN TO LOAD THEN PRESS RETURN";NORMAL$
256 INPUT A$
260 PRINT HOME$+CENTER$+INVERSE$+*LOADING *+PG$+NORMAL$
270 GOSUB 10000\ GOTO 30
1000 CHAIN "LOAD1.BAS" WITH HOME$, INVERSE$, NORMAL$, FLASH$, CENTER$
2000 CHAIN "LOAD2.BAS" WITH HOME$, INVERSE$, NORMAL$, FLASH$, CENTER$
3000 CHAIN "LOAD3.BAS" WITH HOME$, INVERSE$, NORMAL$, FLASH$, CENTER$
4000 CHAIN "LOAD3A.BAS" WITH HOME$, INVERSE$, NORMAL$, FLASH$, CENTER$
5000 CHAIN "LOAD4.BAS" WITH HOME$, INVERSE$, NORMAL$, FLASH$, CENTER$
6000 PRINT HOME$;CENTER$;INVERSE$;"RETURNING TO MENU";NORMAL$
6005 CHAIN "MENU, BAS"
6010 END
10000 REM LOAD PROGRAM INTO DOWN HOLE INSTRUMENT SBR.
10100 INPUT #1%,A$
10150 OPEN "XK:" AS FILE #2% REM COMMUNICATIONS PORT
10200 FOR N=6 TO LEN(A$) STEP 3
10300 VAR$=MID$(A$,N,2)\ IF VAR$="ZZ" THEN CLOSE #1%\ GOTO 11500
10500 FOR I1=2 TO 1 STEP -1\NUM$=MID$(VAR$, I1, 1)
10510 IF NLM$ <= "9" THEN NLMBER=VAL (NLM$) ELSE NLMBER=ASCII (NLM$)-55
10515 IF I1=2 THEN TOTAL=NUMBER ELSE IF I1=1 THEN TOTAL=TOTAL+(NUMBER*16)
10550 NEXT 11
10600 PRINT #2%, CHR$(TOTAL);
11460 NEXT N
11470 GOTO 10100
11500 CLOSE #1%,#2%
11510 PRINT HOME$+CHR$(27)+"[12B"+CHR$(27)+"[12C"+FLASH$;
11520 PRINT *PROGRAM TRANSMITTED*+NORMAL$\ PRINT
11530 PRINT INVERSES; SWITCH INSTRUMENT TO RESET ;NORMALS
11550 PRINT "PRESS RETURN TO CONTINUE ":\ INPUT A$
11600 RETURN
```

SET NO DOUBLE 1000 PROGRAM LOAD1(HOME\$, INVERSE\$, NORMAL\$, FLASH\$, CENTER\$) 1001 REM 19 DEC 85 1005 PRINT HOME\$+INVERSE\$+"*** LOAD PARAMETERS ***** +NORMAL\$ 1010 OPEN "XK:" AS FILE #2% REM COMMUNICATIONS PORT 1015 PRINT \ PRINT INVERSE\$+*MOVE INTERFACE SWITCH FROM RESET TO RUN";NORMAL\$ 1017 LINPUT #2% A\$ 1018 N2=0 1019 B**\$=""** 1020 LINPUT #2%,A\$\ IF LEN(A\$)<1 THEN 1020 ELSE B\$=B\$+A\$ 1025 IF ASCHI(MID\$(A\$,LEN(A\$),1))=3 THEN GOSUB 1200\ GOTO 1027 ELSE 1020 1027 PRINT 8\$;"(DECIMAL) ";\N2=N2+1\ 1F N2=4 THEN N2=1 1030 INPUT A\$\DEC=VAL(A\$)\ GOSUB 1500\ FOR I1=1 TO LEN(HEX\$) 1031 PRINT #2%, MID*(HEX*, 11, 1)\ GOSUB 1033\ NEXT 11\ GOSUB 1033 1032 PRINT #2%, CHR\$(13)\ GOTO 1040 1033 FOR 12=1 TO 50\ NEXT 12\ RETURN 1040 IF HEX\$="0000" AND N2=1 THEN 1045 ELSE 1019 1045 N=1 1049 B\$="" 1050 LINPUT #2%,A\$\ IF LEN(A\$)<1 THEN 1050 ELSE B\$=B\$+A\$ 1060 IF ASCII(MID\$(A\$,LEN(A\$),1))=13 THEN C\$(N)=B\$\N=N+1\B\$="" 1070 IF ASCII(MID\$(A\$,LEN(A\$),1))=3 THEN C\$(N)=B\$\ GOTO 1090 1080 GOTO 1050 1090 FOR N1=2 TO N-1 1100 SAM\$=MID\$(C\$(N1),11,4)\INTVL\$=MID\$(C\$(N1),28,4)\DELY\$=MID\$(C\$(N1),42,4) 1110 HEX\$=SAM\$\ GOSUB 1600\ PRINT MID\$(C\$(N1),1,10)+DEC\$; 1120 HEX\$=INTVL\$\ GOSUB 1600\ PRINT MID\$(C\$(N1),15,13)+DEC\$; 1130 HEX\$=DELY\$\ & & SOSUB 1600\ PRINT MID\$(C\$(N1),32,10)+DEC\$ 1140 NEXT N1\ PRINT C\$(N) 1150 CALL INKEY (A\$)\ IF LEN(A\$)=0 THEN 1150 1160 IF ASCII(A\$)=13 THEN PRINT #2%,CHR\$(13)\ GOTO 1017 1170 IF A\$="G" THEN PRINT #2%, "G" 1190 PRINT HOME\$+CENTER\$+INVERSE\$+*INSTRUMENT RUNNING*+NORMAL\$\ PRINT 1190 PRINT *PRESS RETURN FOR MENU* : \ INPUT A\$ 1195 CHAIN "LOADER.BAS" 1200 REM B\$ CLEANUP 1210 IF LEN(B\$))11 THEN B\$=MID\$(B\$,LEN(B\$)-10,11) 1220 RETURN 1500 REM DEC TO HEX CONVERT 1505 HEX\$="" 1510 FOR N=12 TO 4 STEP -4 1515 HEX=0 1520 IF DEC<2^N THEN 1530 ELSE DEC=DEC=2^N\HEX=HEX+1\ GOTO 1520 1530 IF HEX<10 THEN HEX\$=HEX\$+CHR\$(HEX+48) 1540 IF HEX>9 THEN HEX\$=HEX\$+CHR\$(HEX+55) 1550 NEXT N 1560 IF DEC<10 THEN HEX\$=HEX\$+CHR\$(DEC+48) 1570 IF DEC>9 THEN HEX\$=HEX\$+CHR\$(DEC+55) 1580 RETURN 1600 REM HEX TO DEC CONVERT 1610 FOR 11=4 TO 1 STEP -1\W\$=MID\$(HEX\$,11,1) 1620 IF W\$<="9" THEN D1=VAL(W\$) ELSE D1=ASCII(W\$)-55 1630 IF 11=4 THEN DN=D1 1640 IF I1=3 THEN DN=DN+(D1*16) 1650 IF 11=2 THEN DN=DN+(D1*256) 1660 IF I1=1 THEN DN=DN+(D1*4096) 1670 NEXT I1\DEC\$=NLM\$(DN) 1680 RETURN

SET NO DOUBLE 2000 PROGRAM LOAD2(HOME\$, INVERSE\$, NORMAL\$, FLASH\$, CENTER\$) 2001 REM 19 DEC 85 2005 CENTER1\$=CHR\$(27)+"[13;35H" 2010 DIM \$4, DA\$(1400,1)\ PRINT HOME\$+INVERSE\$+"*** DUMP DATA ***"+NORMAL\$ 2015 OPEN "INTERIM" AS FILE #4, VIRTUAL 2016 PRINT CENTER\$; INVERSE\$; "CLEARING INTERIM STORAGE" :NORMAL\$ 2018 FOR N1=0 TO 1400\DA\$(N1,0)=**\DA\$(N1,1)=**\ NEXT N1 2020 OPEN "XK:" AS FILE #2% REM COMMUNICATIONS PORT 2030 PRINT \ PRINT INVERSES; "SWITCH INSTRUMENT CONTROL TO RESET THEN PRESS "; 2040 PRINT "RETURN";NORMAL\$\ INPUT A\$ 2050 LINPUT #2,A\$\ IF LEN(A\$)>0 THEN 2050\ REM CLEAR COMM PORT 2055 IF LEN(A\$)>0 THEN 2050 2060 PRINT \ PRINT "SWITCH INSTRUMENT CONTROL TO ";FLASH\$;"RUN";NORMAL\$;" NOW" 2070 LINPUT #2%,A\$\ IF LEN(A\$)<1 THEN D%=D%+1\ GOTO 2100 ELSE DA\$(N,0)=DA\$(N,0)+A\$ 2080 IF N=0 THEN PRINT HOME\$; CENTER\$; FLASH\$; "DUMPING DATA"; NORMAL\$ 2090 IF LEN(DA\$(N,0)) <6 THEN 2070 ELSE N=N+1\ PRINT CENTER1\$;N\D%=0\ GOTO 2070 2100 IF D%(800 THEN 2070 2110 CLOSE #2% 2120 PRINT HOME\$+CENTER\$+INVERSE\$+*FORMATTING DATA*+NORMAL\$\ GOSUB 2290 2130 CLOSE #2% 2140 PRINT HOME\$ 2150 FOR N1=0 TO NN PRINT DA\$(N1,1), NEXT N1N PRINT 2160 PRINT FLASH\$+*DUMP TO DISK? (N OR RETURN) "+NORMAL\$;\ INPUT A\$ 2170 IF A\$="N" THEN 2280 2180 PRINT "FILENAME ";\ INPUT NA\$ 2190 PRINT HOME\$+CENTER\$+INVERSE\$+"WRITING ";NA\$+NORMAL\$ 2200 OPEN NA\$ FOR OUTPUT AS FILE #3% 2210 PRINT #3%,N 2220 FOR N1=0 TO N 2230 PRINT #3%, DA\$(N1,1) 2240 NEXT N1 2250 CLOSE #3% 2260 PRINT HOME\$+CENTER\$+FLASH\$+"FILE ":NA\$;" STORED"+NORMAL\$\ PRINT 2270 PRINT "PRESS RETURN FOR MENU" ;\ INPUT A\$ 2275 CLOSE 2280 CHAIN "LOADER.BAS" 2290 REM RECONFIGURE DATA FROM D\$ TO DA\$ 2295 D\$=""\N3=0 2300 FOR N1=0 TO N-1 2320 FOR N2=1 TO LEN(DA\$(N1,0))\A\$=MID\$(DA\$(N1,0),N2,1) 2330 IF ASCII(A\$)>47 AND ASCII(A\$)<90 THEN D\$=D\$+A\$ 2335 IF LEN(D\$)=4 THEN DA\$(N3,1)=D\$\N3=N3+1\D\$=""\ PRINT CENTER1\$;N3 2340 NEXT N2\ NEXT N1 2350 N=N3-1\ RETURN

6. LOAD3.BAS

.

```
SET NO DOUBLE
3000 PROGRAM LOAD3(HOME$, INVERSE$, NORMAL$, FLASH$, CENTER$)
3001 REM 19 DEC 85
3005 DIM #4,DA$(1400,1)\ PRINT HOME$+INVERSE$+**** DISPLAY FILE DATA ***** HNORMAL$
3006 OPEN "INTERIM" AS FILE #4, VIRTUAL
3007 PRINT CENTER$; INVERSE$; *CLEARING INTERIM STORAGE*; NORMAL$
3008 FOR N=0 TO 1400\DA$(N,0)=""\NEXT N
3010 GOSUB 9000
3020 FOR N1=0 TO NN PRINT USING "#### 'LLL ",N1,DA$(N1,0);
3030 IF N1-INT(N1/84)*84=83 THEN 3035 ELSE 3040
3035 PRINT \ PRINT *PRESS RETURN TO CONTINUE*; \ INPUT A$\ PRINT HOME$
3040 NEXT NI\ PRINT \ PRINT "DISPLAY DATA AGAIN? (Y OR RETURN)":\ INPUT A$
3045 IF AS="Y" THEN PRINT HOMES \ GOTO 3020
3048 CLOSE
3050 CHAIN "LOADER.BAS"
9000 PRINT \ PRINT "FILENAME ";\ INPUT NA$
9005 PRINT HOME$+CENTER$+INVERSE$+*READING * (NA$+NORMAL$
9010 OPEN NA$ FOR INPUT AS FILE #2%
9030 INPUT #2%, NN REM NUMBER OF RECORDS IN FILE
9040 FOR N1=0 TO N
9050 INPUT #2%,DA$(N1,0)
9060 NEXT N1
9070 CLOSE #2%
9075 PRINT HOME$
9080 RETURN
```

7. LOAD3A.BAS

SET NO DOUBLE 4000 PROGRAM LOAD3A(HOME\$, INVERSE\$, NORMAL\$, FLASH\$, CENTER\$) 4001 REM 19 DEC 85 4002 DIM #4,DA\$(1400,1) 4003 OPEN "INTERIM" AS FILE #4, VIRTUAL 4005 PRINT HOME\$+INVERSE\$+***** PRINT FILE DATA *****HORMAL\$ 4006 PRINT CENTER\$; INVERSE\$; *CLEARING INTERIM STORAGE*; NORMAL\$ 4007 FOR N=0 TO 1400\DA\$(N,0)=""\ NEXT N 4010 GOSUB 9000 4015 PRINT HOME\$; CENTER\$; INVERSE\$; "PRINTING "; NA\$; NORMAL\$ 4020 OPEN *LP:* FOR OUTPUT AS FILE #22/\ REM SET LINE PRINTER OUTPUT 4035 PRINT #2%, NA\$\ PRINT #2% 4040 FOR N1=0 TO N STEP 6 4050 FOR N2=0 TO 5 4055 PRINT #2% USING "#### 'LLLL *, INT(N1+N2), DA\$(N1+N2,0); 4060 NEXT N2\ PRINT #2%\ NEXT N1 4070 CLOSE 4080 CHAIN "LOADER.BAS" 3000 PRINT \ PRINT "FILENAME ";\ INPUT NA\$ 9005 PRINT HOME\$+CENTER\$+INVERSE\$+*READING *;NA\$+NORMAL\$ 3010 OPEN NA\$ FOR INPUT AS FILE #2% 9030 INPUT #2%,N\ REM_NUMBER OF RECORDS IN FILE 9040 FOR N1=0 TO N 3050 INPUT #2%,DA\$(N1,0) 9060 NEXT N1 9070 CLOSE #2% 9075 PRINT HOME\$ 9080 RETURN

LOAD4.BAS 8. 5000 PROGRAM LOAD4(HOME\$, INVERSE\$, NORMAL\$, FLASH\$, CENTER\$) 5001 REM 19 DEC 85 5002 DECLARE DOUBLE A, B, R1, R2, AL, BE, GA, TM, LR\DIM #4, DA\$(1400,1) 5003 OPEN "INTERIM" AS FILE #4, VIRTUAL 5005 PRINT HOME\$+INVERSE\$+**** CONVERT & PRINT FILE DATA ***** +NORMAL\$ 5008 FOR N=0 TO 1400\DA\$(N,0)=""\ NEXT N 5010 PRINT "INSTRUMENT NUMBER? ";\ INPUT A\$\I=VAL(A\$) 5015 RESTORE 5020 FOR N=0 TO IN READ A, B, R1, AL, BE, GAN NEXT N

5030 PRINT "CRUISE NUMBER? ";\ INPUT CR\$\ PRINT "DATE? ";\ INPUT DA\$ 5040 60SUB 9000\N2=N 5045 OPEN NA\$+".TXT" FOR OUTPUT AS FILE \$1% 5050 PRINT HOMES; CENTERS; INVERSES; "CONVERTING & PRINTING"; NORMALS 5060 DATA .14797245,.55253495,8715,8.9910055E-4,2.4362808E-4,1.2626278E-7 5062 DATA .14797245,.55253495,8715,8.9910055E-4,2.4362808E-4,1.2626278E-7 5064 DATA .14797245,.55253495,8715,8.9910055E-4,2.4362808E-4,1.2626278E-7 5066 DATA .14797245,.55253495,8715,8.9910055E-4,2.4362808E-4,1.2626278E-7 5068 DATA .14797245..55253495.8715.8.9910055E-4.2.4362808E-4.1.2626278E-7 5070 DATA .14797245,.55253495,8715,8.74722104E-4,2.47238861E-4,1.16103772E-7 5072 DATA .14797245,.55253495,8715,8.84648075E-4,2.45607753E-4,1.21657048E-7 5074 DATA .14797245, .55253495, 8715, 8.78525336E-4, 2.46894380E-4, 1.17247016E-7 5076 DATA .14797245,.55253495,8715,8.83512916E-4,2.46027504E-4,1.17903032E-7 5078 DATA .14797245,.55253495,8715,8.50827418E-4,2.50936456E-4,1.03111000E-7 5080 DATA .14797245,.55253495,8715,8.78246674E-4,2.46585729E-4,1.17843303E-7 5100 OPEN "LP:" FOR OUTPUT AS FILE #2% REM SET UP LINE PRINTER 5105 L%=0\ REM LINE COUNTER 5110 PRINT #2%, "CRUISE # ";CR\$;" DATE ";DA\$\L%=L%+1 DATE ;DA\$ 5115 PRINT #1%, "CRUISE # ";CR\$;" 5120 PRINT #2%, "INSTRUMENT # ";I\L%=L%+1 5125 PRINT #1%, "INSTRUMENT # ";I 5130 REM PRINT CHR\$ (27); CHR\$ (58); 5140 PRINT #2%,NA\$ 5145 PRINT #1%,NA\$+".DAT" TEMP"\L%=L%+3 5150 PRINT #2% PRINT #2%, * REC COUNT RESISTANCE 5155 PRINT #1% PRINT #1%," REC COUNT RESISTANCE TEMP* 5160 FOR I=0 TO N2 5170 FOR I1=4 TO 1 STEP -1\C\$=MID\$(DA\$(I,0),I1,1) 5180 IF I1=1 THEN IF ((C\$>"3") AND (C\$<"8")) OR (C\$>"B") THEN DN=9999 5190 IF I1=1 THEN 5260 5200 IF C\$(="9" THEN D1=VAL(C\$)\ GOTO 5220 5210 D1=ASCII(C\$)-55 5220 IF I1=4 THEN DN=D1 5230 IF I1=3 THEN DN=DN+(D1*16) 5240 IF I1=2 THEN DN=DN+(D1*256) 5250 GOTO 5270 5260 IF C\$ ("8" THEN DN=-DN 5270 NEXT 11 5280 PRINT #2% USING "####",I; 5285 PRINT #1% USING "####",I; 5290 IF MID\$(DA\$(I,0),1,4)="STOP" THEN PRINT #2%," STOP";\ PRINT #1%," STOP";\GOT0 5360 5300 IF DN=9999 THEN PRINT #2%," OVER";\ PRINT #1%," OVER";\GOTO 5360 5310 PRINT #2% USING "######## , DN;\PRINT #1% USING "######## , DN;

5330 LR=LOG(R2)\TM=(1/(AL+LR*BE+LR^3*GA)-273.15)

5340 PRINT #2% USING " #####.### TM\L%=L%+1\PRINT #1% USING " #####.### .TM

5345 IF L%=60 THEN L%=0\ PRINT #2%,CHR\$(12) 5360 NEXT I 5370 PRINT #2%,CHR\$(12) 5380 CLOSE #2%\CLOSE #1% 5385 CLOSE" 5390 CHAIN "LOADER.BAS" 9000 PRINT \ PRINT "FILENAME ";\ INPUT NA\$ 9005 PRINT HOME\$+CENTER\$+INVERSE\$+"READING ";NA\$+NORMAL\$ 9010 OPEN NA\$ FOR INPUT AS FILE \$2% 9030 INPUT #2%,N\ REM NUMBER OF RECORDS IN FILE 9040 FOR N1=0 TO N 9050 INPUT #2%,DA\$(N1,0) 3060 NEXT N1 9070 CLOSE #2% 9075 PRINT HOME\$ 9080 RETURN

APPENDIX H. COMPUTER PROGRAMS TO REDUCE HPC TEMPERATURE DATA

LANGUAGE: FORTRAN-4

MACHINE: VAX 11/780 (Virtual Memory System)

<u>INTRODUCTION</u> - Two programs (DECAY1, FITTING1) have been developed to extrapolate temperature measured over time after penetration of a core barrel (HPC) into bottom sediments. These are presently written as separate programs, although they could be combined as subroutines called by a main program.

<u>REFERENCE</u>: K. Horai, A theory of processing downhole temperature data taken by the hydraulic piston corer (HPC) of the DSDP, ms. in preparation.

DECAY1 - Computes theoretical cooling temperature of an idealized core barrel in sediment, for the normalized initial temperatures of 1°C for the HPC, and O°C for the sediment. The core barrel is assumed to be a perfect conductor (uniform temperature) although with finite heat capacity, with 2-dimensional (radial) conduction to the sediment.

Instructions for running the program from a computer terminal

- 1. You will be asked by the terminal for values of a and b; the inner and outer radii, respectively, of the HPC. The values of a and b (units of meters) are typed in from the terminal.
- Give a 6-digit file name to identify the cooling curve. If you give <u>uvwxyz</u>, the file name to be created will be <u>Tuvwxyz</u>.<u>DAT</u>.
- 3. Then you will be asked for the thermal conductivity of the sediment. Type in its value (units of W/mK).
- 4. Then you will be asked for the time interval between data points. Check the temperature record to be compared with the theory, and give it in the units of seconds.
- 5. The program computes 3500 terms of the integrand given in the theory (these are displayed on the terminal as computations progress) and 200 cooling temperatures at the specified time interval by numerically integrating the integrands, multiplied by an exponentially decaying term as given in the theory. The results are stored in the file.

FITTING1 - Compares the theoretical cooling curve with the temperature record to determine the optimal fit by least-squares analysis.

Instructions to run the program:

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1. You will be asked for the 6-digit thermal decay file name. Type in from the terminal keyboard <u>Tuvwxyz.DAT</u> file.

- 2. Then you will be asked for a 6-digit file name to specify the corresponding data. Type in from the terminal <u>Dabcdef.DAT</u> to open this data file.
- 3. Type in the number (IO) of "bad" (poorly fitting) data points (the first part of the cooling curve) that are not used in the analysis. Sometimes this must be done iteratively (trial and error) for noisy or otherwise uncertain data.
- 4. The printout will give the residual fit for various shift (NO) in origin of data points vs. theoretical curve. This is done to determine an effective origin time for uncertain penetrations. The data and the theoretical cooling temperatures will be printed for the value of NO that gives the minimum residual.

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5. Abstract (Limit: 200 words)

A miniature temperature recorder has been developed to be used with the hydraulic piston sediment corer (HPC) on the Deep Sea Drilling Project (DSDP). The instrumentation fits into pressure-sealed slots in the wall of the HPC, allowing temperature measurements to be made simultaneously with coring operations. Temperatures from -2 to 70° C are measured to a resolution of about 0.01° C. Up to 1300 13-bit measurements are recorded in random access memory (RAM), at a sampling rate ranging between 0.1 s to over 100 min., as specified by the operator in a program loaded into a microprocessor of the instrument. During recording the instrumentation uses about 3.5 mamp at 7.5 volts, which can be supplied for about 20 hours of operation by a custom-made pack of silver-oxide batteries. The corer is normally left motionless in the sediment for about 10 min. to allow extrapolation of the measured temperatures to equilibrium in situ temperature. Examples of data from DSDP Leg 86 are given.

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